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Influence of Dielectric Loading on the Fidelity Factor of an Ultra Wideband Monopole Antenna

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Abstract—**This paper reports a study of the influence on the fidelity factor due to loading an Ultra Wideband monopole with a pair of dielectric substrate pads. The fidelity factor was calculated, with and without loading, using the simulated transmitted and received pulses in three points in the azimuth plane. Results show that this dielectric loading technique does not cause significant pulse distortion. The fidelity factor is affected only by an average of 2.8% relative to the unloaded antennas in all situations considered. Two versions of the same antenna design were studied: coplanar- and microstrip-fed. It was found that the influence of feeding on the fidelity factor is also minimal.**

Keywords-UWB monopole antenna; dielectric loading; fidelity factor; microstrip–fed; CPW–fed

I. INTRODUCTION

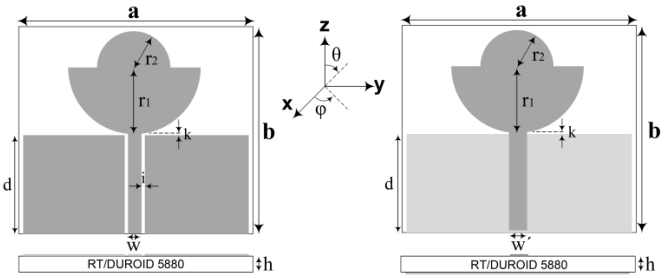
The demand for Ultra–wide Band (UWB) antennas has required significant development efforts to achieve small designs for a variety of applications such as consumer devices, wireless body area network, radar, biomedical imaging, and localization [1–6].

The effects on UWB monopoles of a dielectric loading sandwich technique, consisting of gluing two commercial substrate pads on both sides of an antenna, have been previously studied and presented [1–2]. It was found that this technique increases the antenna electrical size and improves matching resilience without significant efficiency decrease. Other possible antenna performance trade-offs need to be evaluated, therefore, this paper aims at studying the impact on the fidelity factor of this technique.

Effects on the time performance of dielectric loading are, to the best of the authors' knowledge, still unreported. The most used parameter to assess time performance is the antenna fidelity factor [7–8]. In our study, the fidelity factor was calculated with and without the sandwich padding. The same UWB monopole base design was considered with two different common feeding structures: microstrip– (MS–fed) and coplanar–fed (CPW–fed) in order to also evaluate the influence of the feeding structure on fidelity.

II. ANTENNA DESIGN AND STRUCTURE

Two identical printed UWB antennas with different feeding structures were designed and simulated. Their configuration and dimensional parameters are shown in Fig. 1. In both versions, the radiator patch comprises two semicircles with different radii. The backside of the CPW–fed antenna substrate is devoid of any metallization while the backside of the MS–fed antenna supports a finite ground plane. The two designs differ in the feeding line width (w for CPW and w' for MS) that was calculated to have an input impedance of 50 Ω in both structures. The used substrate was RT/Duroid^{TM} 5880 with thickness of 1.57 mm, relative permittivity of 2.2 and loss tangent of 0.0009. To achieve the desired antenna performance, *e.g.* wider impedance characteristic, the dimensions of the proposed antenna must be optimized [3]. The final dimensions are presented in Table 1. Among the dimensional parameters shown in Table I, the performance of both antennas is mainly affected by the feed gap (k), the radiator shape (r_1, r_2) and the ground plane size.



(a) (b)

Figure 1. Printed UWB monopole antennas: (a) CPW-fed, (b) MS-fed

TABLE I. OPTIMAL PARAMETER VALUES OF BOTH STRUCTURES IN MM

a	*b*	r_1	r_2	*d*	*w*	*k*	*i*	*h*	*w'*
44	38	11	7.5	18	3.2	0.2	0.15	1.57	4.9

The dielectric loading sandwich technique was applied to both antennas adequately choosing pad size and permittivity. Two identical rectangular slices of de–metalized commercial substrate (RO3003^{TM}) with a permittivity of 3.0, loss tangent of 0.001 and thickness of 1.52 mm were padded on each side of the two structures, as seen in Fig. 2 for the CPW-fed antenna prototype. The width and length of the dielectric pads are 19 and 22 mm, respectively. It should be noted that larger loading pads cause a higher loading effect, *i.e.,* the antenna has enhanced

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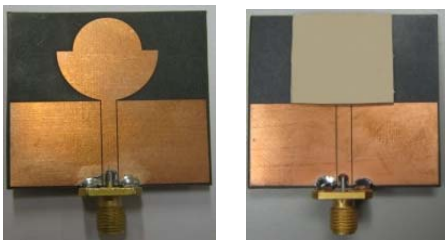
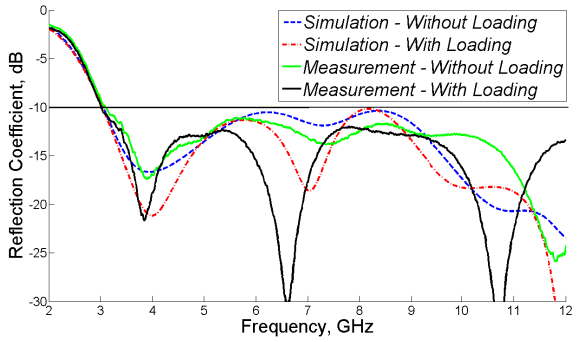
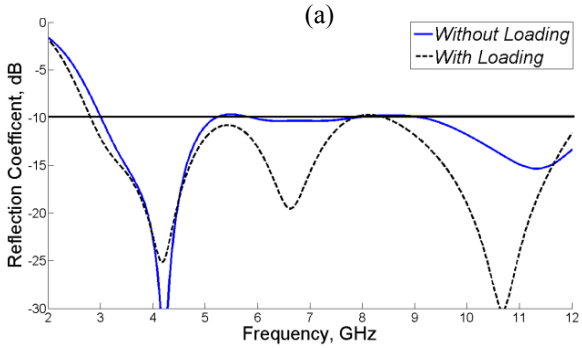


(a) (b)

Figure 2. CPW-fed built prototype: (a) without loading, (b) with loading



(a)



(b)

Figure 3. Simulated reflection coefficient results of antenna: (a) CPW-fed, (b) MS-fed

matching level, although lower radiation efficiency is expected. The loaded antennas dimensions were not further adjusted taking the sandwich into account in order to have a fair performance comparison in the time domain between the loaded and unloaded versions.

III. Results and Discussion

Results for both frequency and time domain are reported in this section. Simulations have been carried out using the commercial software package CST Microwave Studio™ 2012.

A. Frequency Domain

Fig. 3 illustrates the reflection coefficient of the designed antennas with and without loading. Experimental results are also presented for the CPW–fed prototype (Fig. 2) that was built and measured to confirm the validity of the simulation model with and without loading.

Both unloaded antennas have an input impedance bandwidth extending in the range 3 to 12 GHz, fulfill the requirement of VSWR < 2 (equivalent to $S_{11} < -9.5$ dB), and cover more than the FCC–defined frequency band.

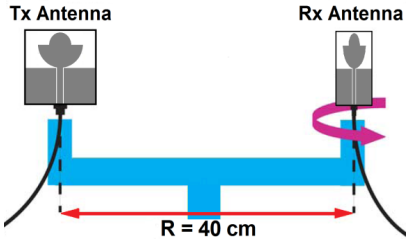


Figure 4. Simulation setup

For the loaded antennas the operating bandwidth is mainly affected in the lower frequencies.

B. Time Domain

The time-domain simulation setup is depicted in Fig. 4. It uses two identical antennas for transmitting (Tx) and receiving (Rx). The Tx antenna is fixed and is facing the Rx antenna ($\varphi = 0°$, $\theta = 90°$). The Rx antenna rotates in the azimuth plane ($\varphi = 0°$, 90°, and 180°, $\theta = 90°$). The chosen distance between the antennas is 40 cm, which is four times the wavelength at the lower operating frequency (3 GHz), large enough to validate the far-field approximation.

A Gaussian modulated sine pulse (CST default excitation) was used as the source pulse with spectrum corresponding to the 3.1 GHz – 10.6 GHz frequency range. Figs. 5 and 6 illustrate the impulse responses for both antennas with and without loading. The results show good transient performance of both antennas at different orientations. In the plots, the received signals (Rx) with and without loading are similar; hence, the technique does not impose significant pulse distortion. a well–known parameter, the *fidelity factor* [7], has been used to quantify the pulse distortion level. It basically compares the shapes of a source pulse, $S_t(t)$, and a received pulse, $S_r(t)$, and it has been defined by the Equation below (see [9], for instance).

$$\max_{\tau} \int_{-\infty}^{+\infty} \frac{S_t(t)}{|S_t(t)|^2} \cdot \frac{S_r(t-\tau)}{|S_r(t)|^2}\, dt$$

The fidelity is a parameter that estimates the maximum correlation coefficient of the two signals by varying the time delay τ. In fact, it reflects the similarity between the transmitted and received pulses. The fidelity factor of the considered structures was calculated by using the Rx and Tx pulses of Fig. 5 and 6 and the results are shown in Table II. For the studied antennas the fidelity factor is minimally affected by the feeding structure with an average deviation of 1.8% of the CPW–fed antenna to the MS–fed antenna for every orientation. The simulation results show that dielectric loading does not significantly impact fidelity factor of both antennas with a degradation of only 2.8% on average and relative to the unloaded antennas in all situations considered. The worst case, with a 10% decrease in the parameter, occurs at $\varphi = 90°$ for both structures. This is due to the different radiation patterns of the loaded *vs.* unloaded antennas. Unloaded antennas have higher directivity in $\varphi = 90°$ and loading causes a decrease in the same direction. The fidelity factor in $\varphi = 90°$ is therefore higher for the unloaded antennas and lower when loading is applied. It should be noted that both structures have lower gain in the 9 GHz region leading to an overall reduction of the

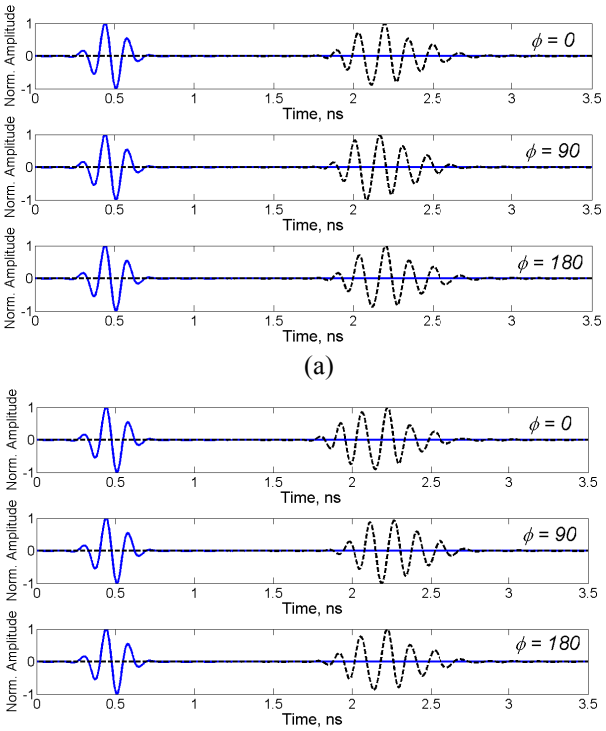


Figure 5. Tx and Rx Signal of the Antenna without Loading: (a) CPW-fed, (b) MS-fed

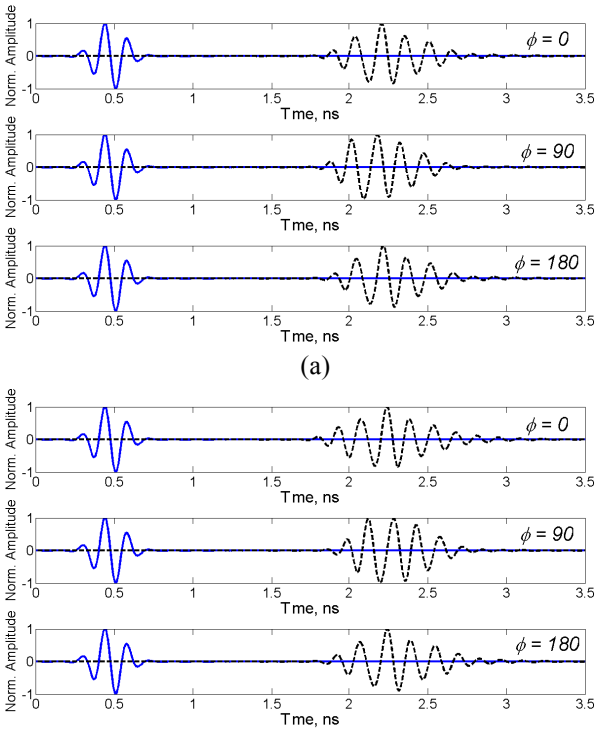


Figure 6. Tx and Rx Signal of the Antenna with Loading: (a) CPW-fed, (b) MS-fed

fidelity factor values. The parameter has an average of about 0.84 in this study, which is higher than the commonly accepted minimum of 0.5 [8].

IV. CONCLUSION

A study of the influence of dielectric loading on the fidelity factor of an UWB printed monopole antenna has been presented. The following observations can be made: proper loading does not impose significant distortion on the transmitted pulses which leads to an average of 2.8% deviation in fidelity factor from the unloaded antennas; it has been confirmed that adequate loading does not significantly detune the antennas.

This study also aimed at evaluating the importance of feeding structure. To this goal, two versions of the same UWB printed monopole antenna have been studied: CPW– and MS–fed. Antenna input matching has been simulated and measured for one of the designs to confirm the model accuracy. It was found that the antenna feeding structure has minimal impact on the fidelity factor with a 1.8% average deviation from CPW– to MS–fed.

TABLE II. FIDELITY FACTOR

Antenna Type	*Phi (deg)*	*Fidelity Factor*	
		Without Loading	*With Loading*
CPW-fed	***φ = 0***	0.83	0.83
	φ = 90	0.86	0.81
	φ = 180	0.83	0.83
MS-fed	***φ = 0***	0.83	0.82
	φ = 90	0.88	0.79
	φ = 180	0.81	0.81

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