Real-time Delay with Network Coding and Feedback

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Abstract

We consider the problem of minimizing delay when broadcasting over erasure channels with feedback. A sender wishes to communicate the same set of μ messages to several receivers. The sender can broadcast a single message or a combination (encoding) of messages to all receivers at each timestep, through separate erasure channels. Receivers provide feedback as to whether the transmission was received. If, at some time step, a receiver cannot identify a new message, delay is incurred. Our notion of delay is motivated by real-time applications that request progressively refined input, such as the successive refinement of an image encoded using multiple description coding.

Our setup is novel because it combines coding techniques with feedback information to the end of minimizing delay. Uncoded scheduling or use of erasure correction coding, such as maximum distance separable (MDS) codes, has been well-studied in the literature. We show that our setup allows $\Theta(\mu)$ benefits as compared to both previous approaches for offline algorithms, while feedback allows online algorithms to achieve smaller delay compared to online algorithms without feedback.

Our main complexity results are that the offline minimization problem is NP-hard when the sender only schedules single messages and that the general problem remains NP-hard even when coding is allowed. However we show that coding does offer complexity gains by exhibiting specific classes of erasure instances that become trivial under coding schemes. We also discuss online heuristics and evaluate their performance through simulations.

Keywords: broadcasting, feedback, delay, erasures, linear coding, computational complexity, online scheduling and coding algorithms.

1 Introduction

Current and emerging applications, such as satellite imaging, roadside to vehicle communication, internet tv, wireless downlink broadcasting, require content to be downloaded quickly and reliably from a host over possibly unknown channels. In practical networks, transmissions are subject to errors: packets get dropped due to congested links, wireless fading and interference, expired timestamps, etc. Such losses are perceived as packet erasures at higher layers, and are often modeled using independent erasure channels.

To cope with unknown channels, feedback information is often available at the broadcasting source. Thus the source, when deciding what to transmit next, knows which subset of receivers successfully received each of its past transmissions. That is, the source has perfect feedback information. Feedback can be efficiently employed in a wireless environment: the source might acquire such information by taking advantage of the symmetry of wireless links, or by collecting acknowledgment packets explicitly using specifically designed control traffic [10], or implicitly, by overhearing transmissions from the receiver nodes [16]. In satellite transmissions, a satellite might learn when a receiver goes in a deep fade (e.g., enters a tunnel), in which case it loses a sequence of packets. It is also known how to explicitly collect acknowledgments in wired networks, when the source multicasts the same content over a distribution tree in an overlay network [13].

In this paper, we consider the problem of combining coding techniques and feedback information over broadcasting channels to offer reliable content delivery under delay guarantees. Our notion of delay is motivated from real-time applications with progressively refined input. Such a paradigm is provided by multiple description coding that we adopt as our illustrating example in the following; however, our notion of delay is relevant to a much more general class of applications.

Multiple description is a well studied data compression technique which allows to provide robustness and graceful recovery in the presence of unknown channel conditions. Although the theoretical problem was already introduced in the 80's (see for example [14]), the research interest in the field was significantly invigorated during the last few years, due to the numerous identified network applications, such as image and video delivery (see for example [25, 9, 3, 26, 27], and [15] for a tutorial paper). The main idea is that we encode our file, for example an image, using a number μ of equally important descriptions, and each description is send separately to the receiver. Depending on the channel conditions, the receiver may receive a different number of descriptions. These descriptions are constructed to have the following property: if a receiver receives a single description (any one of them), it gets a coarse version of the image that is within some appropriately defined distortion guarantees from the original. If a receiver gets any two descriptions, it can reconstruct the image more accurately. Generally, the larger the number of descriptions, the smaller the distortion distance from the original. Reception of all descriptions results in the most accurate reconstruction. Note that in this construction, it is only the number of different received files that defines the reconstruction accuracy; the ordering at which files are received plays no role.

Consider now an application that requires fast delivery of images over a wireless network, for example from a road-basestation of a transportation network to passing vehicles. Assume that the image is encoded using multiple description, and thus the basestation has μ blocks to deliver. When communicating towards a single receiver, simple sequential transmission of the blocks suffices: the underlying multiple description coding will determine the image quality experienced by the receiver, as a function of the number different blocks collected. The problem becomes much more challenging when the image needs to be broadcasted to a number of receivers, each of which receives information over its own erasure channel. The sender may use a scheduling algorithm to decide which image block to broadcast next. In this paper, we propose instead to use a *coding* algorithm, that encode the blocks we need to transmit to the receivers. Both in the case of scheduling or coding, the algorithm may use the feedback information it has collected (i.e., which receivers received the previous transmissions) to decide on the current transmission. Note that our proposed coding is additional to the multiple description data compression: it decides which and how many image blocks it will combine together, and falls in the area of network coding, as its main purpose is to better share the network resources among the contending receivers. Network coding is an emerging area that has attracted a very significant research interest during the last decade, as it promises to revolutionize information flow through networks (see [1, 18] and [12] for an introductory tutorial).

Every time receiver r_j receives successfully, it wants to learn some missing piece of information, namely any image description it does not know yet. This motivates us to increment the delay d_j of r_j by one every time r_j successfully receives a transmission of the following type: (i) an image description r_j already knows, or (ii) an encoding of image descriptions which, when combined with r_j 's successful receptions so far, does not allow r_j to immediately extract at least one image block r_j does not know yet. This definition allows us to disengage delay from the erasure frequency as we only count delay when a transmission is successful. It also allows us to capture two causes of delay: delay due to useless received packets, namely packets that bring duplicate information to their receiver, and delay due to packets that, although useful, do not allow their receiver to immediately decode some unknown message. Finally, our definition of delay is the simplest instantiation possible, as it does not take into account any ordering: we thus hope that a good understanding of this problem can serve as a first step towards more combinatorially demanding delay definitions.

The main questions we consider in this paper are (i) whether coding offers benefits in terms of delay, and (ii) how to design coding schemes that minimize average and maximum delay, and what is the complexity of this task. We focus in the case where all receivers are interested in the same content

because understanding this simple model offers a first step towards variations, where receivers may demand different subsets of the messages or request the messages in a specific order. It is worth noting that the popular solution of employing rate-less erasure correcting codes at the source such as LT or Raptor codes [20, 28] for reliable broadcasting over erasure channels, performs very poorly in terms of delay (see also Subsection 1.1).

Our contributions include the following. Concerning the complexity of the offline problem, we show that minimizing the average and maximum delay when the source uses scheduling is NP-hard. We then examine the complexity of the problem when coding is allowed and show that, although specific classes of erasure instances become trivial, the general problem remains NP-hard. We examine classes of erasure instances where coding offers significant benefits in terms of delay, and give a simple inapproximability result for maximum offline delay. Finally, we discuss heuristic online algorithms where the erasures of different receivers are independent and i.i.d. distributed. We evaluate the performance of our heuristics through simulations. The latter verify our observation that coding can significantly reduce delay compared to scheduling.

The importance of our work lies perhaps in that, to the best of our knowledge, it is the first to examine the complexity and algorithmic aspects of the joint use of coding and feedback information for delay-optimal content delivery. Erasures are inherent in many realistic networks, and we believe that the trade-off beween rate and delay that arises in our setting is worth exploring further.

The remainder of this paper is organized as follows. Section 2 introduces our model and notation. Section 3 examines the complexity of offline broadcasting with scheduling, while Section 4 examines the complexity when coding at the source is allowed. Section 5 discusses online results. Section 6 concludes.

1.1 Related Work

A significant body of work has investigated the problem of scheduling user requests over a broadcast medium to maximize the per-user received rate and minimize the response time; see for example [7, 4, 8, 22]. In this setup, users typically arrive at different time instances, and ask for possibly different content. No coding is employed and no errors are assumed. The difficulty of the problem, which was recently shown to be NP-hard [7], arises from having to share the common medium over the contenting requests. It is worth mentioning that if receivers were to ask for the same data items (e.g., satellite images) in any order, then even if the requests arrive at different time instances, a periodic (circular) transmission of the data items would suffice.

In uncoded transmissions, maximizing the per user throughput naturally minimizes the delay: the faster information is received, the better. However, in the presence of erasures, uncoded transmission leads to repetitive reception and cannot achieve rates close to the optimal, and thus can also not achieve the optimal delay. On the other hand, when coding is employed, delay and rate may become conflicting requirements. For rate-less codes for example, to operate close to capacity and avoid duplicate receptions, we need to encode at the source across μ packets, for large values of μ [20, 28]. A receiver needs to wait to collect $\Theta(\mu)$ coded packets to be able to decode, which implies a delay of $\Theta(\mu)$. Indeed, in the presence of erasures, satisfying requests even for the same content becomes a challenging problem [11]. In [?, 29] use of MDS codes has also been proposed, but their performance is inferior to Raptor codes both in terms of complexity and adaptability to unknown channel conditions.

Our work can also be viewed as an instantiation of network coding with feedback. Recent work has looked at use of acknowledgments and coding to optimize the achievable rate, under the condition that each received packet is either useless or can be immediately decoded by the destination [16]. Such schemes, although simple to implement, do not offer rate or delay guarantees. Another line of work has looked into use of coding and feedback to minimize the queue size at the sender [30]. This performance metric is quite different from delay. Also, [23] examines use of feedback over broadcast erasure channels to optimize rate and achieve zero probability of error.

Finally, Birk and Kol [6] introduced a related broadcasting scenario, called Index Coding. Index

Coding was examined in a line of works [5, 19, 2]. A more general setting than Index Coding, recently analyzed by Alon et. al. in [2], is as follows. A source possesses m messages. There are n receivers. Each receiver somehow, e.g., through some initial phase where the source schedules single messages only and every receiver receives these messages through an erasure channel, ends up with a subset of the m messages by some time t; this subset is the side information of the receiver. It is then assumed that no receiver experiences any erasures after time t, and that each receiver wants exactly one of the source messages it is still missing. The goal is to find the minimum length of the codeword whose transmission will allow all receivers to simultaneously recover the messages they want.

Our problem differs from the formulation above in many ways. First, we do not restrict our model to guarantee that there is some time t after which every receiver will have a perfect communication channel with the source. Second, the source is allowed to transmit linear combinations at any time slot, so even if there was such a time t, the side information of a receiver would not just be single messages. More importantly, given that we do not impose any time t after which every receiver successfully receives the source transmissions, it is not difficult to see that applying at every time step an erasure correcting code that allows every successfully receiving receiver to reconstruct one of its missing message does not necessarily minimize delay.

As far as we know, this is the first work that examines jointly optimizing coding and use of feedback information for delay-optimal content delivery. Our paper builds on a preliminary work [17], where we introduced the problem and proposed some online heuristics.

2 The Model

Consider a source that wants to convey μ messages to ρ receivers using broadcast transmissions. Time is slotted, and at the beginning of each time slot $t \geq 1$, the source is allowed to transmit (linear or non-linear) combinations of the messages, which we call packets. We denote the packet transmitted at the beginning of time slot $t \geq 1$ by p(t). We call a broadcasting scheme that schedules single (uncoded) messages at every time step a *scheduling* scheme. If the scheme is allowed to use coding operations (combinations) on the messages, we call it a *coding* scheme. Each receiver receives information from the source through an erasure channel, which might range from only deep fadings to i.i.d. erasures and may depend on the erasure channels of other receivers. We denote by $K_j^t \in \{0,1\}$ the realization of receiver r_j 's channel at time t with $K_j^t = 1$ if and only if r_j receives p(t). In a worst case model these realizations could have given values, while in a probabilistic setting they would be random variables.

Depending on whether the transmitted packet successfully reached a receiver or not, the receiver sends an ACK or a NACK to the source respectively (we assume that the feedback channels are perfect). We assume that K_j^t is received by the source at the end of time slot t. Therefore the source can use this information for generating the packet p(t+1) transmitted at the next time slot t+1. We assume that the duration of the time slot is sufficient for the receivers to receive the packet and decode it (if possible) using the packets they have already received. A receiver who has decoded all μ messages is no longer interested in the source transmissions. The source transmits a packet during every time slot until all receivers have decoded all messages.

We can think of the μ source messages as defining a μ -dimensional space over a finite field \mathbb{F}_q , where each message corresponds to one of the orthonormal basis vectors $\{e_1, e_2, \dots, e_{\mu}\}$. We will denote by p(t) the linear packet 1 the source transmits at time t. Linear packets are of the form (c, x) where $c \in \mathbf{F}_q^{\mu}$ and $x = \sum_j c_j e_j$; the choice of the coefficient vector c determines x, so we leave x implied in what follows. Operations over a finite field \mathbf{F}_q of size say $q = 2^{\ell}$ in practice means that we divide the binary packets the source produces into contiguous sets of ℓ bits, and treat each such set as a symbol of \mathbf{F}_q . Linear combining of the packets occurs symbol-wise.

Let Π_j^t be the subspace collected by r_j at the end of time slot t and E_j^t the set of vectors $e_\ell \in \Pi_j^t$.

It is not difficult to generalize this discussion to packets that consist of nonlinear combinations of messages.

We say that a received vector (packet) brings novel information to a receiver r_j if it increases the dimension of Π_j by one. A class of schemes that will play an important role henceforth are schemes where every successfully received packet brings innovative information to its receiver. We call these schemes rate-optimal. In a rate-optimal scheme, a receiver r_j that has received ℓ packets, has collected an ℓ -dimensional subspace Π_j of the μ -dimensional space. For $\ell = \mu$, the receiver can successfully decode all source messages. The following properties of rate-optimal schemes are straightforward (also see [17]):

- 1. A receiver r_i can decode the source message e_i if and only if $e_i \in \Pi_i$; and
- 2. With slight abuse of notation, let Π_1 denote the subspace spanned by the vectors $\langle p(i_1), \ldots, p(i_{t_1}) \rangle$ receiver r_j has collected by time t_1 (recall that r_j may have some erasures during the t_1 time steps, hence $i_{t_1} \leq t_1$), and Π_2 the subspace spanned by the vectors $\langle p(i_{t_1}+1), \ldots, p(i_{t_2}) \rangle$ the same receiver r_j collects between times t_1+1 and t_2 . If E_1 is the set of vectors $e_{\ell} \in \Pi_1$ and E_2 the set of vectors $e_{\ell} \in \Pi_2$, then $E_1 \cap E_2 = \emptyset$, for all j, $t_1 < \mu$ and $t_2 \leq \mu$.

Let $1(\cdot)$ be the indicator function.

Definition 1 The delay d_j^T that receiver r_j experiences when the source uses transmission scheme T is the number of packets that, although successfully received, did not allow r_j to immediately decode a new message. ² In symbols,

$$d_j^T \triangleq 1 + \sum_{t:|E_j^t| < \mu} \mathbf{1}(E_j^t = E_j^{t-1}) \cdot K_j^t$$

Let D_a^T and D_w^T denote the average and worst case delay of transmission scheme T respectively, given by

$$\mathcal{D}_a^T \triangleq rac{1}{
ho} \sum_{1 \leq j \leq
ho} d_j^T, \quad ext{and} \quad \mathcal{D}_w^T \triangleq \max_{1 \leq j \leq
ho} d_j^T.$$

Different schemes may result in different values for \mathcal{D}_a^T and \mathcal{D}_w^T . Our goal is to find among all transmission schemes the (possibly different, e.g., see [17]) transmission schemes under which the average and worst-case delay are minimized.

As an example, given our definition of delay, if a given broadcasting instance allows for the minimum delay of one, the scheme must be rate-optimal or proved rate-optimal because of the realization of the channels. In other words, may have transmitted packets that were not novel for certain receivers during certain time steps but the specific receivers experienced erasures at the specific time steps. On the other hand, if a scheme is not rate optimal, then any successfully received packet that does not bring new information to its receiver, increments the delay of its receiver by one. Hence any scheme that is not rate optimal given the realizations of the channels will have average delay strictly larger than one and maximum delay at least two.

The discussion above implies that further insight into the online problem can be gained by examining its offline version. The offline broadcasting instance has two additional inputs: (a) an integer τ that stands for the number of source transmissions by which all receivers have received all messages, and (b) a $\tau \times \rho$ symbolic matrix P whose entries take values from $\{\sqrt{x}\}$. P is defined by the successful receptions and the erasures of the ρ receivers during the τ time slots, which are known in the offline scenario: entry $P(t,j) = \sqrt{1}$ if and only if receiver r_j successfully received the packet p(t) transmitted at time slot t. We shall henceforth denote an offline broadcasting instance by the quadruple (μ, ρ, τ, P) , and refer to P as the erasure matrix of the instance.

We say that a broadcasting scheme for the source *completes* the offline instance (μ, ρ, τ, P) if by time τ , all receivers have decoded all messages. Observe that a necessary and sufficient condition for a scheme to complete the offline instance is that every receiver has at least μ successful receptions by time τ . For

²We introduce the +1 in the delay for technical reasons –we can interpret this as setup time: e.g., the time slot t = 0 is used by the source to identify the number of receivers in the system.

³We introduce P(t,j) for the offline scenario while we use K_i^t in the online scenario.

example, any rate-optimal scheme requires exactly μ successful receptions per receiver to guarantee that all receivers can decode all messages, regardless of the delay.

Offline analysis is useful because it can be used as a benchmark for the online algorithms: the performance of the optimal offline algorithm lower bounds the performance of *any* online algorithm. Moreover, offline problems can be particularly interesting and challenging on their own, as the works on Index Coding [6, 5, 19, 2] show.

3 Minimizing Delay with Scheduling Schemes is NP-hard

Given an offline broadcasting instance (μ, ρ, τ, P) , the *scheduling* problem we are considering is to minimize the average (maximum) delay under any scheduling scheme that completes the instance. Observe that a priori this appears to be an easier problem than the one studied in [7] since our notion of delay is relaxed as all receivers need all messages and not just specific subsets of messages, and further, the order in which the messages are received does not matter.

The decision version of the optimization problem above has as an extra input an integer $d \ge 1$, and answers "yes" if and only if there is a *scheduling* scheme that completes (μ, ρ, τ, P) with total (maximum) delay at most d^4 . An algorithm that solves the minimization problem for total (maximum) delay should be able to answer the decision problem for every value of d. Since the minimum possible value for both average and maximum delay is one (the delay corresponding to the initialization phase), and since average delay of one implies maximum delay of one, it suffices to prove that it is hard to decide if the average delay is one, in order to prove that both minimization problems are NP-hard. This is the main result of this section and it is summarized in the following theorem.

Theorem 1 Minimizing average and maximum delay in offline broadcasting in the presence of erasures and when the source uses scheduling schemes is NP-hard.

In the rest we will prove Theorem 1 by reducing 3SAT to average delay of one in offline broadcasting. We shall henceforth refer to average delay simply as delay.

Given a formula ϕ in CNF on n variables x_1, \ldots, x_n , and m clauses c_1, \ldots, c_m , where each clause consists of disjunctions of exactly 3 literals, we want to decide if there exists an assignment of truth values to the variables such that all clauses are satisfied.

We will construct an offline broadcasting instance $B(\phi) = (\mu, \rho, \tau, P(\phi))$ such that ϕ is satisfiable if and only if there is a scheduling scheme that completes $B(\phi)$ with delay one. In our instance, the source has $\mu = 2n$ messages, there are $\rho = n + 2m$ receivers, and $\tau = 4n + 5m$ time slots. Our construction guarantees that each receiver has exactly $\mu = 2n$ successful receptions by time τ . Notice that this choice of τ suffices to decide if there is a delay-one scheme for our instance: any such scheme has to be rate-optimal and therefore must deliver all μ messages to every receiver during the first μ successful receptions of the receiver starting at t=1.

In more detail, our construction works as follows. For every variable x_i , $1 \le i \le n$ we introduce 2 messages, e_i and $\overline{e_i}$. One receiver D^i is introduced for every variable x_i (their role will be discussed after the construction of $P(\phi)$ is complete). Also, two receivers, C_1^j and C_2^j are introduced for every clause c_j , $1 \le j \le m$. This results in the erasure matrix $P(\phi)$ having $\rho = n + 2m$ columns.

We now move on to discussing the number of rows in $P(\phi)$. For every variable x_i , we introduce 4 consecutive time slots, which we call the variable period β_i ; β_i starts at time slot 4(i-1)+1, and ends at time slot 4i. Following the n-th variable period, we introduce m consecutive clause periods: the j-th clause period, denoted by γ_j , consists of 5 time slots, starts at time slot 4n + 5(j-1) + 1, and ends at time slot 4n + 5j. Hence $P(\phi)$ has $\tau = 4n + 5m$ rows.

To complete our construction, we need to assign values to the $\tau \cdot \rho$ entries of $P(\phi)$. We will do this sequentially in time, i.e., by first considering the *variable* periods and then the *clause* periods.

 $^{^{4}}$ Minimizing total delay is equivalent to minimizing average delay when m is independent of n.

Time slot	C_1^j	C_2^j
4(i-1)+1		X
4(i-1)+2	X	X
4(i-1)+3	X	$\sqrt{}$
4i	X	X

Time slot	C_1^j	C_2^j
4(i-1)+1		X
2(i-1)+2	X	X
4(i-1)+3	X	X
4i	X	

Time slot	C_1^j	C_2^j
4(i-1)+1		
4(i-1)+2		
4(i-1)+3	X	X
4i	X	X

Table 1: Erasure patterns for receivers C_1^j , C_2^j during β_i . If clause c_j contains x_i , they receive as in the left table; if c_j contains \overline{x}_i , they receive as in the middle table; else $(c_j$ does not contain x_i or \overline{x}_i), they receive as in the right table.

Time slot	D^1	 D^{i-1}	D^i	D^{i+1}	 D^n
4(i-1)+1		 	X		
4(i-1)+2		 $\sqrt{}$	X	$\sqrt{}$	
4(i-1)+3	X	 X	$\sqrt{}$	X	 X
4i	X	 X	$\sqrt{}$	X	 X

Time slot	C_1^j	C_2^j
4n + 5(j-1) + 1		
4n + 5(j-1) + 2	X	
4n + 5(j-1) + 3	X	
4n + 5(j-1) + 4		X
4n + 5j		X

Table 2: The left table shows receptions of D^1, \ldots, D^n during β_i . The right table shows receptions of C_1^j, C_2^j during clause period γ_j (all other receivers experience erasures during γ_j).

During variable period β_i , for all $1 \leq j \leq m$, receivers C_1^j, C_2^j corresponding to clause c_j receive as shown in Table 1 depending on whether $x_i, \overline{x_i}$ or none of them appears in c_j . Also, during β_i , receivers D^ℓ for $1 \leq \ell \leq n$, receive as shown in Table 2: for $\ell \neq i$, D^ℓ receives during the first two time slots of β_i , while D^i receive during the last two time slots.

During clause period γ_j , receivers C_1^j , C_2^j corresponding to clause c_j receive as shown in the right table of Table 2. All other receivers experience erasures during γ_j .

The above completes our construction. Table 4 in the Appendix shows $P(\phi)$ for the example formula $\phi = (x_1 \vee x_2 \vee x_3) \wedge (\overline{x}_1 \vee \overline{x}_2 \vee x_3) \wedge (\overline{x}_2 \vee \overline{x}_3 \vee x_4)$ for which $B(\phi) = (8, 10, 31, P(\phi))$.

Some remarks are appropriate at this point. First it is trivial to check that the reduction can be carried out by a deterministic Turing machine in logarithmic space, and that every receiver has exactly μ successful receptions. So a priori there could be a scheduling scheme completing $B(\phi)$ with delay one. The role of the receivers D^i is to guarantee that exactly 2 messages are sent during each β_i , with the two messages sent during the first two time slots being rescheduled during the last two time slots, in any order (see Proposition 2 for a proof). In effect, this flexibility in the scheduling of the messages during the last two time slots of each β_i is our choice gadget. Our consistency gadget is that during β_i , C_2^j receives a different message from C_2^ℓ when x_i appears in clause c_j and \overline{x}_i in c_ℓ . Finally our clause constraint gadget is the simultaneous reception of the two receivers corresponding to clause c_j during the first time slot of γ_i .

We now move to showing that ϕ is satisfiable if and only if $B(\phi)$ admits delay one. Before, we introduce the following two schedulings which will prove useful for our arguments.

Scheduling 1 for variable period β_i : the ordered sequence of messages $e_i, \overline{e}_i, e_i, \overline{e}_i$.

Scheduling 2 for variable period β_i : the ordered sequence of messages $e_i, \overline{e}_i, \overline{e}_i, e_i$.

Proposition 1 If ϕ is satisfiable, then there is a scheduling scheme T_S that satisfies the offline broadcasting instance $B(\phi) = (2n, 2m + n, 4n + 5m, P(\phi))$ with delay one.

Proof. Consider a satisfying truth assignment for ϕ . For $1 \le i \le n$, if x_i is true, T_S applies Scheduling 1 for β_i during β_i , else if x_i is false, T_S applies Scheduling 2 for β_i during β_i . Then the first 4n transmissions of T_S incur delay one, and D^1, \ldots, D^n obtain all messages.

Since ϕ is satisfiable, every clause has at least one literal that is true. W.l.o.g., let $c_j = (\ell_i \vee \ell_a \vee \ell_b)$ be any clause, where ℓ_y is either x_y or \overline{x}_y , and suppose that ℓ_i is (one of) the satisfying literal(s) for this clause, i.e., x_i is set to true if and only $\ell_i = x_i$. We now show how T_S completes the clause periods so that the clause receivers obtain all messages without delay.

By time 4n, receivers C_1^j , C_2^j know 2n-6 messages, i.e., all messages corresponding to the variables that do not appear in clause c_j . Further (see Table 1), C_1^j knows $\{e_i, e_a, e_b\}$, and C_2^j knows exactly one from $\{e_a, \overline{e}_a\}$ and one from $\{e_b, \overline{e}_b\}$. C_2^j also knows e_i which he received at the third time slot of β_i if x_i appears in c_j (in which case, since ℓ_i is the satisfying literal of c_j , x_i was set to true, and T_S applied Scheduling 1 for β_i during β_i), or at the fourth time slot of β_i if \overline{x}_i appears in c_j (in which case, x_i was set to false and T_S applied Scheduling 2). Then during the first time slot of γ_j (see Table 2), T_S sends \overline{e}_i . Next, T_S sends e_a or \overline{e}_a , depending on which one C_2^j missed during β_a ; similarly for e_b during the third time slot. Finally T_S schedules \overline{e}_a and \overline{e}_b during the last two slots of γ_j . Since these transmissions result in C_1^j , C_2^j obtaining all messages without delay, T_S satisfies $B(\phi)$ with delay one.

Conversely, let T_S' be any scheduling scheme that satisfies $B(\phi)$ with delay one. We will exhibit a satisfying truth assignment for ϕ . We first introduce some notation: for $1 \leq t_1 \leq t_T \leq 4$, we define $E_i^{t_1...t_T}$ to be the (simple) set of the messages scheduled at the discrete time steps t_1, \ldots, t_T of β_i . For example, E_i^{123} is the simple set of messages sent during the first 3 slots of β_i . We can now show a technical but useful lemma concerning properties of T_S' during the variable periods.

Proposition 2 Consider any scheduling scheme T'_S that satisfies $B(\phi)$ with delay one. For $1 \le i \le n$, T'_S schedules exactly two new messages during β_i , with the messages sent during the first two time slots of β_i being resent (in some order) during the last two time slots. In symbols, for all $1 \le i \ne j \le n$, $E_i^{12} = E_i^{34}$ and $E_i^{12} \cap E_j^{12} = \emptyset$.

Proof. Trivially, T_S' is rate-optimal for every D^i since it satisfies $B(\phi)$ with delay one. Thus $|E_i^{12} \cup E_i^{34}| \geq 2$ for all i. Now suppose that there is a k such that $E_k^{12} \neq E_k^{34}$. It follows that $|E_k^{12} \cup E_k^{34}| \geq 3$, so by the pigeonhole principle, there is a message m that was scheduled during both β_k and some β_j with $j \neq k$. During β_k , m was either received by all D^ℓ with $\ell \neq k$ or by D^k or both. During β_j , m was either received by all D^i with $\ell \neq k$ or by $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ and all $\ell \neq k$ or by $\ell \neq k$ or b

W.l.o.g., assume that T_S' schedules the two messages e_{x_i}, e_{y_i} during the first two time slots of β_i , in this order. By Proposition 2, these messages will not be rescheduled before time 4n, so for the sake of clarity, we may relabel them as e_i, \overline{e}_i respectively. We define the following truth assignment. For $1 \leq i \leq n$, if T_S' applied Scheduling 1 for β_i during β_i , x_i is set to true, else if T_S' used Scheduling 2 for β_i during β_i , x_i is set to false. Notice that Proposition 2 guarantees that any T_S' indeed applied one of these two schedulings during β_i .

We are now ready to conclude the converse direction of our reduction after stating one more proposition whose proof will appear shortly.

Proposition 3 Let $c_j = (\ell_i \vee \ell_a \vee \ell_b)$ be any clause. Any T'_S that satisfies $B(\phi)$ with delay one is such that C_2^j has received at least one of e_i, e_a, e_b by time 4n.

Corollary 1 If T'_S is a scheduling scheme that satisfies $B(\phi)$ with delay one then ϕ is satisfiable.

Proof. Consider any clause $c_j = (\ell_i \vee \ell_a \vee \ell_b)$. By Proposition 3, any T_S' is such that C_2^j has received at least one of e_i, e_a, e_b by time 4n. W.l.o.g., assume C_2^j received e_i . If C_2^j received this message at time 4(i-1)+3, then x_i appears in c_j and T_S' used Scheduling 1 for β_i . Hence our truth assignment set x_i to true. Otherwise, if C_2^j received e_i at time 4i, then \overline{x}_i appears in c_j and T_S' used Scheduling 2 for β_i . Hence our truth assignment set x_i to false. In either case, our truth assignment for x_i satisfies c_j . Since Proposition 3 applies to all C_2^j for $1 \leq j \leq m$, there is (at least) one literal in every clause that is set to true by our truth assignment. Hence ϕ is satisfiable.

We now give the proof of Proposition 3.

Proof. Consider the 2 receivers C_1^j, C_2^j corresponding to c_j . By Proposition 2, under any T_S' , at the beginning of γ_j , each of C_1^j, C_2^j knows the 2n-6 messages that correspond to the n-3 variables that do not appear in c_j . Further, C_1^j knows $\{e_i, e_a, e_b\}$, and C_2^j knows exactly one of $\{e_i, \overline{e}_i\}$, one of $\{e_a, \overline{e}_a\}$, and one of $\{e_b, \overline{e}_b\}$.

Suppose that C_2^j received $\{\overline{e}_i, \overline{e}_a, \overline{e}_b\}$. Then at the first slot of γ_j where he receives simultaneously with C_1^j , there is no way to avoid delay since every message needed by one receiver incurs delay to the other. However if C_2^j had received at least one of e_i, e_a, e_b , say e_i , then at the first time slot of γ_j , T_S' could schedule \overline{e}_i which does not delay any receiver.

4 Benefits and Limits of Coding in Reducing Complexity

We here start by attempting to understand what are structural properties of instances where use of offline scheduling results in delay greater than one. We will then show that use of coding across messages can offer two benefits: (i) Reduce the delay. For example, we will see that there are instances where with coding we can have delay one, while with scheduling we cannot. (ii) Reduces the complexity of solving the problem for several cases. For example, for the erasure pattern in Section 3, we can trivially achieve average and maximum delay one: during β_i , with $1 \le i \le n$, send $e_i, \overline{e}_i, e_i, \overline{e}_i$, while during β_j for clause c_j , with $1 \le j \le m$, send $e_i + \overline{e}_i$, then whatever is missing from C_2^j , and finally $\overline{m}_a, \overline{m}_b$. The main prupose of this section is to examine whether and how much use of coding can help.

We will use the following notation. Let B_t denote the set of messages the source has transmitted up to time t and \overline{B}_t the set of remaining messages. For receiver r_j , let E_j^t denote the set of messages from B_t that r_j has received, and \overline{E}_j^t the messages from B_t it has not. That is, $B_t = E_j^t \cup \overline{E}_j^t$ for all r_j .

For the case of one receiver, trivially, scheduling achieves delay one. For the case of two receivers, we can use the following simple algorithm to ensure delay one: if at time time t (i) both r_1 and r_2 receive, transmit a message from \overline{B}_t (ii) only r_j receives, if $\overline{E}_j^t \neq \emptyset$ transmit a message from \overline{E}_j^t , otherwise a message from \overline{B}_t . This scheme ensures that at each time t either $\overline{E}_1^t = \emptyset$ or $\overline{E}_2^t = \emptyset$; moreover, $\overline{B}_t = \emptyset$ only when at least one of the two receivers has received all messages.

For the case of three receivers, offline scheduling can result in worst case delay of $\mathcal{O}(\mu)$. Indeed, note that for scheduling, delay is introduced only when the transmission scheme cannot be rate optimal. For the erasure pattern in Table 3, assume that each line is repeated for $\mu/2$ time slots. Rate optimality for r_3 implies that at $t = \mu + 1$, $\overline{E}_1^t \cap \overline{E}_2^t = \emptyset$ and thus, the transmissions at time-slots $t = \mu + 1, \ldots, 3\mu/2$ will incur sum delay $\mu/2$ for r_1 and r_2 . The existence of receiver r_3 is necessary to ensure that $\overline{E}_1^t \cap \overline{E}_2^t = \emptyset$ occurs in offline.

The following straightforward proposition formalizes this observation.

Proposition 4 If at time t there exist receivers r_i and r_j such that $\overline{E}_i^t \cap \overline{E}_j^t = \emptyset$, and following time t, for the next D timeslots that r_i successfully receives so does r_j , with $D \triangleq \min\{|\overline{E}_i^t|, |\overline{E}_j^t|\}$, then offline scheduling results in delay $\mathcal{O}(D)$.

time-slots	r_1	r_2	r_3
$t=1,\ldots,\mu/2$		X	
$t = \mu/2 + 1, \dots, \mu$	X		
$t = \mu + 1, \dots, 3\mu/2$			X

time-slots	r_1	r_2	r_3	r_4
$t=1,\ldots,\mu/2$		X		X
$t = \mu/2 + 1, \dots, \mu$	X	$\sqrt{}$		X
$t = \mu + 1, \dots, 3\mu/2$			X	

Table 3: The left table gives an erasure pattern where scheduling incurs delay $\mathcal{O}(\mu)$ but coding achieves delay one; the right table gives an erasure pattern where coding as well incurs delay $\mathcal{O}(\mu)$.

Use of coding allows to make the source transmissions rate optimal, for all possible erasure patterns. For example, for the pattern in the left Table 3, it is sufficient at time-slots $t=\mu+1,\ldots,3\mu/2$ to transmit $\mu/2$ messages from $\overline{E}_1^t + \overline{E}_2^t$. However, in this case delay is introduced, if a receiver cannot decode a received linear combination. This is the case for the pattern in the right Table 3 (see also [17]). It is easy to see that, at time $t=\mu+1$, $\overline{E}_1^t \cap \overline{E}_2^t = \emptyset$, and additionally, $\overline{E}_1^t \subset \overline{E}_4^t$, $\overline{E}_2^t \subset \overline{E}_4^t$. To be rate optimal with respect to r_1 and r_2 we need, like before, to transmit from $\overline{E}_1^t + \overline{E}_2^t$. However, these transmissions cannot be decoded by r_4 . Thus similarly to before we now have:

Proposition 5 If at time t there exist receivers r_i , r_j and r_k such that $\overline{E}_i^t \cap \overline{E}_j^t = \emptyset$, $\overline{E}_i^t \subset \overline{E}_k^t$, $\overline{E}_j^t \subset \overline{E}_k^t$, and following time t, for the next D timeslots that r_i successfully receives so does r_j and r_k , with $D \triangleq \min\{|\overline{E}_i^t|, |\overline{E}_j^t|\}$, then offline coding results in delay $\mathcal{O}(D)$.

Clearly, coding allows to achieve delay one for a larger set than scheduling. Some additional such patterns are described in the following proposition (the proof can be found in the appendix).

Proposition 6 With coding we can achieve delay one when we have an arbitrary number of receivers ρ and: 1. Erasure patterns where each broadcast transmission is successfully received by at most two receivers (this corresponds to high erasure probability). 2. Patterns where each broadcast transmission is not received by at most one receiver (this corresponds low erasure probability).

4.1 Minimizing Delay with Coding Schemes is NP-hard

Given that there exist instances where the problem becomes simpler with the use of coding, the next question is, whether the general problem, when we are allowed to use coding, becomes polynomial time, or remains NP-hard. Note that the problem of maximizing the throughput when multicasting over graphs becomes polynomial time if coding at intermediate network nodes is allowed [?, 21], while, if coding is not allowed, it is NP-hard. However the theorem below shows that this is not the case in our problem.

Theorem 2 Minimizing average and maximum delay in offline broadcasting in the presence of erasures and when the source uses (linear or nonlinear) coding is NP-hard.

The proof of this theorem appears in the Appendix due to space limitations and builds on the ideas in the proof of Theorem 1.

We conclude this section with an inapproximability result following the definitions in [24] (the proof appears in the Appendix).

Proposition 7 Unless P = NP, there is no ϵ -factor approximation (coding) algorithm for maximum delay in offline broadcasting with erasures for $\epsilon < 1/2$.

5 Online Algorithms

In this section, we discuss the competitive ratio of a natural class of systematic rate-optimal online algorithms for minimizing average (maximum) delay in the cases of arbitrary and i.i.d. erasures. We also suggest an online heuristic that improves significantly on the performance of the best heuristic from [17] for i.i.d. erasures.

A systematic coding algorithm uses the first μ transmissions to send all messages once uncoded and then starts sending combinations of messages. Rate optimal systematic algorithms have smaller average delay than their non-systematic variants where linear combinations of all messages are used for every transmission. The competitive ratio of systematic rate-optimal algorithms in the presence of a deterministic adversary who only knows that the source is using such an algorithm is given by the following proposition (a proof appears in the Appendix). The adversary is allowed to incur an erasure at any channel during any time step but is not allowed to eavesdrop any channel.

Proposition 8 For $\mu = O(\rho)$ and arbitrary erasures, a systematic rate-optimal online algorithm is $(\mu - O(1))$ -competitive for minimizing average delay and $(\mu - 1)$ -competitive for minimizing maximum delay.

Proposition 8 motivates us to look at algorithms that are not rate-optimal in the online scenario. Specifically, we examine the case where all ρ channels experience i.i.d. erasures with common constant erasure probability q. Figure 1 compares the performance of scheduling, two heuristics from [17], and a new heuristic algorithm that improves on the latter. A short description of these algorithms appears in the Appendix. As mentioned in Section 1, use of coding is critical to achieve lower delay as simple scheduling performs very poorly. The rate optimal algorithm (systematic FEC) achieves the expected delay of $\mu \cdot q$, which is significantly worse than the performance of our heuristic as q, ρ (graphs below) and μ (simulations not shown here) increase. Hence the interdependence between delay from useless packets and delay from non instantly decodable packets has to be exploited in order to improve the performance of the online algorithm.

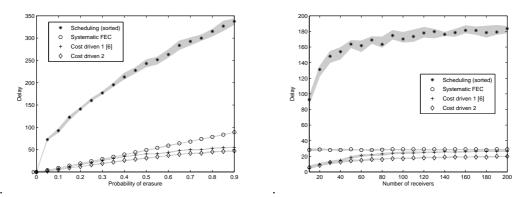


Figure 1: Median delay; the gray area corresponds to confidence interval; $\rho = 150$ receivers on the left graph, erasure probability q = 0.3 on the right graph; $\mu = 100$ messages for both graphs

6 Conclusions

We address the question of finding the optimal offline algorithm for broadcast scheduling or coding that minimizes average and maximum delay when feedback is available. We show that the general problem is NP-hard both when the source uses scheduling and coding. However we show that coding is a reasonable alternative to scheduling as it can significantly reduce delay and alleviate the complexity of scheduling in certain instances. We also suggest online heuristics for minimizing average delay in the presence of i.i.d. erasures with the same erasure probability q across receivers.

The most important open questions at this point are the competitive analysis for more general online algorithms as well as the design of (additive) PTAS for the offline problem.

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Appendix

Time slot	C_1^1	C_2^1	C_1^2	C_2^2	C_{1}^{3}	C_{2}^{3}	D^1	D^2	D^3	D^4
1		X		X		$\sqrt{}$	X			$\sqrt{}$
2	X	X	X	X			X			
3	X	$\sqrt{}$	X	X	X	X		X	X	X
4	X	X	X	$\sqrt{}$	X	X		X	X	X
5		X		X		X		X		$\sqrt{}$
6	X	X	X	X	X	X		X		$\sqrt{}$
7	X	$\sqrt{}$	X	X	X	X	X		X	X
8	X	X	X	$\sqrt{}$	X	$\sqrt{}$	X		X	X
9		X		X		X			X	$\sqrt{}$
10	X	X	X	X	X	X			X	$\sqrt{}$
11	X	$\sqrt{}$	X	$\sqrt{}$	X	X	X	X	$\sqrt{}$	X
12	X	X	X	X	X	$\sqrt{}$	X	X	$\sqrt{}$	X
13		$\sqrt{}$		$\sqrt{}$		X				X
14		$\sqrt{}$		$\sqrt{}$	X	X				X
15	X	X	X	X	X	$\sqrt{}$	X	X	X	$\sqrt{}$
16	X	X	X	X	X	X	X	X	X	$\sqrt{}$
17		$\sqrt{}$	X	X	X	X	X	X	X	X
18	X	$\sqrt{}$	X	X	X	X	X	X	X	X
19	X	$\sqrt{}$	X	X	X	X	X	X	X	X
20		X	X	X	X	X	X	X	X	X
21		X	X	X	X	X	X	X	X	X
22	X	X		$\sqrt{}$	X	X	X	X	X	X
23	X	X	X	$\sqrt{}$	X	X	X	X	X	X
24	X	X	X	$\sqrt{}$	X	X	X	X	X	X
25	X	X	$\sqrt{}$	X	X	X	X	X	X	X
26	X	X		X	X	X	X	X	X	X
27	X	X	X	X		$\sqrt{}$	X	X	X	X
28	X	X	X	X	X		X	X	X	X
29	X	X	X	X	X	$\sqrt{}$	X	X	X	X
30	X	X	X	X	$\sqrt{}$	X	X	X	X	X
31	X	X	X	X		X	X	X	X	X

Table 4: $P(\phi)$ for the example formula $\phi = (x_1 \lor x_2 \lor x_3) \land (\overline{x}_1 \lor \overline{x}_2 \lor x_3) \land (\overline{x}_2 \lor \overline{x}_3 \lor x_4)$ and the scheduling construction in Section 3.

Minimizing Delay with Coding Schemes is NP-hard

In this section we present the proof of Theorem 2. Given an offline instance B(m, n, h, P), the coding problem we are considering is to minimize the delay under any coding scheme that satisfies the instance. Again, delay here stands for average delay.

We will construct an offline broadcasting instance $B(\phi) = (\mu, \rho, \tau, P(\phi))$ such that ϕ is satisfiable if and only if there is a coding scheme that satisfies $B(\phi)$ with delay one. In our instance, the source has $\mu = 3n$ messages, there are $\rho = 5n + 3m$ receivers, and $\tau = 10n + 6m - 3$ time slots. Our construction guarantees that each receiver has $\mu = 3n$ successful receptions within these τ time slots.

Time slot	C_1^j	C_2^j	C_3^j
5(i-1)+1	$\sqrt{}$		X
5(i-1)+2		X	X
5(i-1)+3	X		X
5(i-1)+4	X	X	
5i	X	X	X

Time slot	C_1^j	C_2^j	C_3^j
5(i-1)+1			X
5(i-1)+2		X	X
5(i-1)+3	X		X
5(i-1)+4	X	X	X
5i	X	X	

Time slot	C_1^j	C_2^j	C_3^j
5(i-1)+1		$\sqrt{}$	
5(i-1)+2		$\sqrt{}$	
5(i-1)+3			
5(i-1)+4	X	X	X
5i	X	X	X

Table 5: Erausure patterns for receivers C_1^j , C_2^j , C_3^j during β_i . If clause c_j contains x_i , they receive as in the left table; if c_j contains \overline{x}_i , they receive as in the middle table; else $(c_j$ does not contain x_i or \overline{x}_i), they receive as in the right table.

Time slot	D_1^ℓ	D_2^ℓ	D_3^ℓ	D_4^ℓ	D_5^ℓ
5(i-1)+1			X	$\sqrt{}$	
5(i-1)+2			X,	$\sqrt{}$	
5(i-1)+3					
5(i-1)+4	X	X		X	X
5i	X	X		X	X

Time slot	D_1^i	D_2^i	D_3^i	D_4^i	D_5^i
5(i-1)+1		X	X	X	X
5(i-1)+2	$\sqrt{}$		X	X	X
5(i-1)+3	$\sqrt{}$			X	X
5(i-1)+4	X	X			X
5i	X	X			

Table 6: Let $1 \le k \le 5$. During β_i , D_k^{ℓ} receive as shown in the left table if $\ell < i$; the right table shows receptions for D_k^i .

Our construction works as follows. For every variable x_i , we introduce 3 messages, e_i , \overline{e}_i and e'_i . For every clause c_j , $1 \le j \le m$, we introduce three receivers, denoted by C^j_1 , C^j_2 and C^j_3 . For every variable x_i , $1 \le i \le n$, we introduce five receivers, denoted by D^i_k , for $1 \le k \le 5$, whose role will be discussed after completing the construction of the erasure pattern.

Also, for every variable x_i , we introduce 5 consecutive time slots, which we call the variable period β_i ; β_i starts at time slot 5(i-1)+1, and ends at time slot 5i. Following the n-th variable period, we introduce m consecutive clause periods: the j-th clause period, denoted by γ_j , consists of 6 time slots, starts at time slot 5n+6(j-1)+1, and ends at time slot 5n+6j. Finally, following the m clause periods, we introduce 5n-3 time slots, which we call "patching" time slots, because their role is simply to provide sufficient time for the receivers D_k^i to obtain all messages.

We now proceed to giving values to the $\tau \cdot \rho$ entries of the erasure matrix P. We will do this sequentially in time, i.e., by first considering the variable periods, then the clause periods, and finally the "patching" time slots.

During variable period β_i , for all $1 \leq j \leq n$, receivers C_1^j, C_2^j, C_3^j corresponding to clause c_j receive as shown in Table 5 depending on whether $x_i, \overline{x_i}$ or none of them appears in c_j . Also, during β_i , receivers D_k^ℓ with $1 \leq k \leq 5$ receive as shown in Table 6 if $\ell \leq i$; otherwise (if $\ell > i$), they all experience erasures.

During clause period γ_j , receivers C_1^j , C_2^j , C_3^j corresponding to clause c_j receive as shown in Table 7. All other receivers experience erasures during γ_j .

Finally, consider the "patching" time slots. We can think of them as being grouped into n-1 periods of 5 time slots, and one last period of only two time slots. For $1 \le i \le n$, at time 5n+6m+5(i-1)+1, receivers D_4^i, D_5^i receive. At the next time slot 5n+6m+5(i-1)+2, receivers D_2^i, D_5^i receive. For $1 \le i \le n-1$, during the last three time slots of patching period i, all receivers D_k^i with $\ell > i$, $1 \le k \le 5$ receive.

The above completes our construction. Table 8 shows $P(\phi)$ for the example formula $\phi = (x_1 \lor x_2 \lor x_3) \land (\overline{x}_1 \lor \overline{x}_2 \lor x_3) \land (\overline{x}_2 \lor \overline{x}_3 \lor x_4)$ for which $B(\phi) = (12, 29, 55, P(\phi))$.

Again it is easy to check that the reduction can be carried out by a deterministic Turing machine in logarithmic space and that all receivers have exactly 3n successful receptions by time τ , hence any rate-optimal scheme would satisfy $B(\phi)$. Here the role of the receivers D_k^i is twofold (see Lemma 1 for a proof): they guarantee that (a) a single message is scheduled during every time slot of every β_i , and (b) exactly 3 messages are sent during each β_i , with the two messages sent during the first two time slots being rescheduled during the last two time slots, in any order. Similarly to Section 3, this flexibility in the scheduling of the messages during the last two time slots of each β_i is our choice gadget. Our consistency gadget is that during β_i , C_3^j receives a different message from C_3^ℓ when x_i appears in clause c_j and \overline{x}_i in c_ℓ . Finally our clause constraint gadget is the simultaneous reception of the three receivers corresponding to clause c_j during the first time slot of γ_j .

We now move to showing that ϕ is satisfiable if and only if $B(\phi)$ admits delay one. Before, we introduce the following two schedulings that will prove useful for our arguments.

Scheduling 1 for β_i : the ordered sequence of messages $e_i, \overline{e}_i, e'_i, \overline{e}_i, e_i$.

Time slot	C_1^j	C_2^j	C_3^j
5n + 6(j-1) + 1			
5n + 6(j-1) + 2	X	X	$\sqrt{}$
5n + 6(j-1) + 3	X		
5n + 6(j-1) + 4		X	
5n + 6(j-1) + 5	X		
5n + 6j		X	

Table 7: Receiving pattern of C_1^j, C_2^j, C_3^j during clause period γ_j .

Scheduling 2 for β_i : the ordered sequence of messages $e_i, \overline{e}_i, e'_i, e_i, \overline{e}_i$.

Proposition 9 If ϕ is satisfiable, then there is a coding scheme T that satisfies the offline broadcasting instance $B(\phi) = (3n, 3m +$ $5n, 10n + 6m - 3, P(\phi)$) with delay one.

Proof. Consider a satisfying truth assignment for ϕ . For $1 \le i \le n$, if x_i is true, the coding scheme T applies Scheduling 1 for β_i during variable period β_i . Othewise, if x_i is false, it applies Scheduling 2 for β_i during β_i . Then the first 3n transmissions of T incur delay one, and D_1^1, D_2^1 obtain all messages.

Since ϕ is satisfiable, every clause has at least one literal that is true. W.l.o.g., let $c_j = (\ell_i \vee \ell_a \vee \ell_b)$ be any clause, where ℓ_y is either x_y or \overline{x}_y , and suppose that ℓ_i is (one of) the satisfying literal(s) for this clause, i.e., x_i is set to true if and only $\ell_i = x_i$. We

now show how T completes the clause periods so that the clause receivers obtain all messages without delay. By time 5n, receivers C_1^j , C_2^j , C_3^j know 3n-9 messages, i.e., all messages corresponding to the variables that do not appear in clause c_j . Further (see Table 5), C_1^j knows $\{e_i, \overline{e}_i, e_a, \overline{e}_a, e_b, \overline{e}_b\}$, C_2^j knows $\{e_i, e'_i, e_a, e'_a, e_b, e'_b\}$, and C_3^j knows exactly one from $\{e_a, \overline{e}_a\}$ and one from $\{e_b, \overline{e}_b\}$. C_3^j also knows \overline{e}_i which he received at the fourth time slot of β_i if $\ell_i = x_i$ (in which case, x_i was set to true, and T applied Scheduling 1 for β_i during β_i), or at the fifth time slot of β_i if $\ell_i = \overline{x}_i$ (in which case, x_i was set to false and T applied Scheduling 2).

During γ_j (see Table 7), T sends the following packets: first it sends $\overline{e}_i + e'_i$. This results in C_1^j obtaining e'_i , C_2^j obtaining \overline{e}_i , and C_3^j obtaining e_i' . Next it sends e_i . During the third and the fourth slots, if C_3^j received e_a during β_a , it sends $e_a + \overline{e}_a$, and e_a' respectively. Otherwise, if C_3^j received \overline{e}_a during β_a , it sends $\overline{e}_a + e'_a$ and $e_a + e'_a$ respectively. The situation for ℓ_b is identical. Hence by time 5n + 6j, C_1^j , C_2^j and C_3^j have obtained all messages without delay. At time 5n + 6m + 1, every receiver D_k^i except for D_1^1 and D_3^1 still needs some messages. For $1 \le i \le n$, at the first time slot of

the *i*-th patching period, T sends e_i' , which both D_4^i and D_5^i need (see Table 6). At the second slot, it sends $e_i + \overline{e_i}$ if x_i was set to true, or e_i if x_i was set to false; this satisfies both D_2^i and D_5^i . At this point, D_2^i , D_4^i , D_5^i have received all messages. For $1 \le i \le n-1$ 5, during the last three slots of patching period i, T sends e_i , $\overline{e_i}$, e_i' , which receivers D_k^ℓ with $\ell > i$ had missed during β_i . These transmissions result in D_1^{i+1} , D_3^{i+1} obtaining all messages. This explains why the *n*-th patching period consists of only two slots: the only receivers still requiring some messages are D_2^n , D_4^n , D_5^n . We conclude that by the end of the *n*-th patching period all receivers know all messages. Hence T satisfies $B(\phi)$ with delay one.

Conversely, consider any coding scheme T' that satisfies $B(\phi)$ with delay one. We will exhibit a satisfying truth assignment for ϕ . We first extend our previous notation of $E_i^{t_1...t_T}$ to include all time steps $1 \le t_1 \le t_T \le 5$. We now show a technical but useful lemma concerning properties of T' during the variable periods.

Lemma 1 Any coding scheme T' that satisfies $B(\phi)$ with delay one is rate-optimal and sends a single message at every time step jfor $1 \leq j \leq 5n$. Moreover T' sends exactly 3 messages during every β_i , with the messages sent during the first two time slots of β_i being resent (in some order) during the last two time slots of β_i . After β_i , these messages will not be sent again before (potentially) time 5n + 1.

Proof. Trivially, T' must be rate-optimal. Suppose that at time step 5(i-1)+k, for any $1 \le i \le n$ and $1 \le k \le 5$, a (linear or nonlinear) function of at least two messages is sent. Then receiver D_k^i will delay, as he has received nothing so far. For the second part of the lemma, we need to show that $E_i^{12345} \cap E_j^{12345} = \emptyset$ for all $1 \le i \ne j \le n$. Also, we must show that

 $E^{12} = E^{45}$.

 $E_i^{12} = E_i^{45}.$ Since D_1^1 receives during the first three time slots of every β_i , and the scheme is rate-optimal, $E_i^{123} \cap E_j^{123} = \emptyset$ for all $1 \leq i < j \leq n$. Similarly, since D_3^1 receives during the last three time slots of all β_i , and T' is rate-optimal, $E_i^{345} \cap E_j^{345} = \emptyset$ for all $1 \leq i < j \leq n$. Therefore we need show that $E_i^{12} \cap E_j^{45} = \emptyset$ for all $1 \leq i \neq j \leq n$. Equivalently we just need to show that $E_i^{12} = E_i^{45}$. Oberve that, for all i, D_i^i receives $E_i^{45} \cup E_{i+1}^{123} \cup \ldots \cup E_n^{123}$, and T' is rate-optimal. Hence $E_i^{45} \cap E_j^{123} = \emptyset$ for all j > i. Moreover, for example for i = 1, $|E_1^{45} \cup E_1^{123} \cup \ldots \cup E_n^{123}| = 3n - 1$. Hence D_4^1 knows 3n - 1 messages by time 5n. Since T' satisfies $B(\phi)$ and D_1^1 does not receive after time 5n, D_1^1 must know all 3n messages by time 5n. Since D_1^1 and D_4^1 know the same 3n - 3 messages corresponding to $E_1^{23} \cup \ldots \cup E_n^{123}$, it follows that $E_1^{45} \subset E_1^{123}$. Since $E_1^3 \not\subset E_1^{12}$ as D_1^1 receives during all three time steps 1,2,3, and $E_1^3 \not\subset E_1^{45}$ as D_3^3 receives during all three time steps 3,4,5, we conclude that $E_1^{12} = E_1^{45}$. The same argument holds for all i > 1. for all i > 1.

⁵Recall that the last patching period consists of only two time slots.

W.l.o.g., assume that T' schedules the three messages $e_{x_i}, e_{y_i}, e_{z_i}$ during the first three time slots of β_i , in this order. By Lemma 1, these messages will not be rescheduled before time 5n, so for the sake of clarity, we may relabel them as $e_i, \overline{e_i}, e'_i$ respectively. We define the following truth assignment. For $1 \le i \le n$, if T' applied Scheduling 1 for β_i during β_i , x_i is set to true, else if T' used Scheduling 2 for β_i during β_i , x_i is set to false. Notice that Lemma 1 guarantees that any T' applied one of these two schedulings indeed during β_i .

Similarly to Lemma 1, the following proposition presents necessary and sufficient conditions for the clause periods to be completed with delay one.

Proposition 10 Let $c_j = (\ell_i \vee \ell_a \vee \ell_b)$ be any clause. In any scheme T' that satisfies $B(\phi)$ with delay one, C_3^j has received at least one of $\overline{e}_i, \overline{e}_a, \overline{e}_b$ by time 5n.

Before we give the proof of Proposition 10, we show how it concludes the second direction of our reduction.

Corollary 2 If T' is a coding scheme that satisfies $B(\phi)$ with delay one then ϕ is satisfiable.

Proof. Consider any clause $c_j = (\ell_i \vee \ell_a \vee \ell_b)$. By Proposition 10, T' is such that C_3^j has received at least one of \overline{e}_i , \overline{e}_a , \overline{e}_b by time 5n. W.l.o.g., assume C_3^j received \overline{e}_i . If C_3^j received this message at time 5(i-1)+4, then x_i appears in c_j and T' used Scheduling 1 for β_i . Hence our truth assignment set x_i to true. Otherwise, if C_3^j received \overline{e}_i at time 5i, then \overline{x}_i appears in c_j and T' used Scheduling 2 for β_i . Hence our truth assignment set x_i to false. In either case, our truth assignment for x_i satisfies c_j . Since Proposition 10 applies to all C_3^j for $1 \leq j \leq m$, there is (at least) one literal for every clause that is set to true by our truth assignment. Hence ϕ is satisfiable.

We now give the proof of Proposition 10.

Proof. Consider the 3 receivers C_1^j, C_2^j, C_3^j corresponding to c_j . By Lemma 1, under any T', at the beginning of γ_j , each of C_1^j, C_2^j, C_3^j knows the 3n-9 messages that correspond to the n-3 variables that do not appear in c_j . Further, C_1^j knows $\{e_i, \overline{e}_i, e_a, \overline{e}_a, e_b, \overline{e}_b\}$, C_2^j knows $\{e_i, e'_i, e_a, e'_a, e_b, e'_b\}$ and C_3^j knows exactly one of e_i, \overline{e}_i , one of e_a, \overline{e}_a , and one of e_b, \overline{e}_b .

Suppose that C_3^j did not receive any of $\overline{e}_i, \overline{e}_a, \overline{e}_b$. Then he received e_i, e_a, e_b . Now at the first slot of γ_j where C_1^j, C_2^j, C_3^j receive simultaneously, there is no way to avoid delay: C_3^j needs exactly one single message either from $\{\overline{e}_i, \overline{e}_a, \overline{e}_b\}$, or one from $\{e'_i, e'_a, e'_b\}$, while C_1^j needs no single message from the first set and C_2^j needs no single message from the second set.

On the other hand suppose that C_3^j received at least one of \overline{e}_i , \overline{e}_a , \overline{e}_b , w.l.o.g. \overline{e}_i . Then at time 5n + 6(j-1) + 1, a function of \overline{e}_i and exactly one from $\{e'_i, e'_a, e'_b\}$ would suffice to incur zero delay to all 3 receivers.

Proposition 6 With coding we can achieve delay one when we have an arbitrary number of receivers ρ and: 1. Erasure patterns where each broadcast transmission is successfully received by at most two receivers (this corresponds to high erasure probability). 2. Patterns where each broadcast transmission is not received by at most one receiver (this corresponds low erasure probability). **Proof.** For two sets A, B, the notation A + B means $\{a + b : a \in A, b \in B\}$.

1. If two receivers r_i and r_j receive, transmit a message from $\overline{E}_i^t \cap \overline{E}_j^t$ if not empty, otherwise from $\overline{E}_i^t + \overline{E}_j^t$ if not empty, otherwise from \overline{B}_t . If only r_i receives, if $\overline{E}_i^t \neq \emptyset$ transmit from \overline{E}_i^t otherwise transmit a message from \overline{E}_i^t , otherwise a message from \overline{B}_t .

2. Use systematic transmission for μ time slots. After μ timeslots there exists at most one receiver who has not received each message. Continue by transmitting messages from $+_i \overline{E}_i^t$, where in the summation we include all receivers i that receive the transmission (and have not completed reception).

Proposition 7 Unless P = NP, there is no ϵ -factor approximation (coding) algorithm for maximum delay in offline broadcasting with erasures for $\epsilon < 1/2$.

Proof. Let OPT be the optimal algorithm for offline broadcasting in the presence of erasures, and OPT(x) the maximum delay incurred by OPT on instance $x=(\mu,\rho,\tau,P)$. Let T_C be an ϵ -factor approximation algorithm for maximum delay in offline broadcasting with erasures for some $\epsilon < 1/2$. Then for all inputs x to T_C , $T_C(x) \le \frac{1}{1-\epsilon} \cdot OPT(x)$.

Given a formula ϕ in 3CNF, we construct $B(\phi)$ in polynomial time as described in the proof of Theorem 2 and run T_C on input $B(\phi)$. There are two cases: if $T_C(B(\phi)) = 1$, then we know that ϕ is satisfiable. Otherwise, if $T_C(B(\phi)) \ge 2$, then we conclude that

$$2 \leq T_C(B(\phi)) \leq \frac{1}{1-\epsilon} \cdot OPT(B(\phi)) \Rightarrow 2(1-\epsilon) \leq OPT(B(\phi)) \Rightarrow 1 < OPT(B(\phi)).$$

Since the maximum delay of $B(\phi)$ is one if and only if ϕ is not satisfiable, we conclude that ϕ is not satisfiable. Since $B(\phi)$ can be constructed in polynomial time, and T_C runs in polynomial time, we can decide whether ϕ is satisfiable in polynomial time.

Proposition 8 For $\mu = O(\rho)$ and arbitrary erasures, a systematic rate-optimal online algorithm is $(\mu - O(1))$ -competitive for average delay and $(\mu - 1)$ -competitive for maximum delay.

Proof. During the first μ transmissions, the adversary allows receiver r_j to obtain transmission j and causes erasures to all other receivers. After the μ -th transmission the adversary allows all ρ receivers to successfully receive all transmitted packets.

Since exactly one receiver has obtained each message by time μ , the source must transmit linear combinations of all messages starting at time $\mu + 1$ in order to be rate optimal for every receiver. This incurs a total delay of

$$\rho + (\mu - 2) \cdot \rho + (\rho - \mu)$$

during the following μ transmissions. The first term comes from t=0, the second term from time $\mu+1$ up to time $2\mu-2$ where all receivers delay, and the last term from time $2\mu-1$ where receivers $r_{\mu+1},\ldots,r_{\rho}$ delay (at this point, receivers r_1,\ldots,r_{μ} have successfully received μ packets hence can decode for the μ messages and do not delay). It follows that the average delay is given by

$$\mu - \frac{\mu}{\rho} = \mu - O(1),$$

while the maximum delay is $\mu - 1$ (the delay of any receiver r_j with $j > \mu$).

On the other hand, the optimal offline algorithm incurs average and maximum delay of one: it only transmits e_1 during the first μ transmissions, followed by $e_2, \ldots, e_{\mu}, e_1$ during the following μ transmissions. The proposition follows.

Short Description of the Heuristics in Figure 1, Section 5

- 1. Scheduling (sorted): At every time slot, the message that is requested by most receivers is transmitted.
- 2. Systematic FEC [17]: During the first μ transmissions, all μ messages are transmitted once. Then packets with random linear combinations of the messages are transmitted so that every transmitted packet is rate optimal for every receiver.
- 3. Cost driven 2: This is a variant of Cost driven 1 from [17]. During the first μ transmissions, all μ messages are transmitted once. From then on, the packet p(t) is built as follows: first a packet s(t) that is instantly decodable for every receiver is built in a random manner; s(t) might not be novel for every receiver. At this point, the algorithm sets p(t) = s(t) and then augments p(t) by adding more messages to build the final packet as follows. While there are still messages that are not included in p(t) and have not yet been chosen by the algorithm, one such message m is chosen at random. This message is added to p(t) (multiplied by a random coefficient c) according to the following condition. Assume that all receivers will receive p(t). If the number of receivers that will not experience delay upon reception of $p(t) + c \cdot m$ is larger than the number of receivers that will not experience delay upon reception of $p(t) + c \cdot m$. When the erasures are i.i.d. and identically distributed across receivers, the assumption that all receivers will receive p(t) may be replaced by the assumption that the expected number of receivers receive p(t) without affecting the condition for adding m to p(t).

$t \ C_1^1 C_2^1 C_3^1 \ $	$C_1^2 C_2^2 C_3^2$	$C_1^3 C_2^3 C_3^3$	D_1^1	$D_2^1 D_2^1$	D_A^1 I	$D_5^1 I$	O_1^2	$D_2^2 I$	D_{3}^{2} 1	$D_4^2 L$	$D_5^2 D_1^3$	D_2^3	D_3^3	D_{Λ}^{3}	D_5^3	D_1^4	D_2^4	D_3^4	D_4^4	D_5^4
$1 \sqrt{\sqrt{x}}$	$\sqrt{\sqrt{x}}$	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$		X X	X	х	X	X	X	X X	X	X	X	X	X	X	X	X	X	Х
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	$\sqrt{\mathbf{x}}$ x	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$		\sqrt{x}	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
$3 \times \sqrt{x}$	$x \sqrt{x}$	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$		$\sqrt{}$	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
4 x x √	X X X	X X X	X	X		X	X	X	X	X Y		X	X	X	X	X	X	X	X	X
5 x x x	$X X \sqrt{}$	X X X	X	X √	$\sqrt{}$	$\sqrt{\parallel}$	X	Χ :	X	х х	X	X	X	X	X	X	X	X	X	X
$6 \sqrt{\sqrt{x}}$	$\sqrt{\sqrt{X}}$	$\sqrt{\sqrt{X}}$		√ X		$\sqrt{ }$	<u> </u>			х х		X	X	X	X	X	X	X	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	\sqrt{X} X	\sqrt{X} X	\ <u>\</u>	√ X		√ '	<u> </u>	<u> </u>	X	X X		X	X	X	X	X	X	X	X	X
$\begin{vmatrix} 8 & x & \sqrt{x} \\ 9 & x & x & \sqrt{x} \end{vmatrix}$	$\begin{array}{c cccc} X & & X \\ \hline X & X & X \end{array}$	$\begin{array}{c cccc} X & & X \\ \hline X & X & X \end{array}$	X	$\frac{V}{X} \frac{V}{}$		$\frac{V}{X}$	V X	$\frac{V}{X}$	<u>V_</u>	$\frac{X}{\sqrt{\lambda}}$		X	X	X	X	X	X	X	X	X
10 x x x	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	X X X	X	$\frac{x}{x} $			X	X 1	<u>v</u>	$\frac{\sqrt{1}}{\sqrt{1}}$	/ x	X	X	X	X	X	X	X	X	X
$11 \sqrt{\sqrt{x}}$	$\sqrt{\sqrt{x}}$	$\sqrt{\sqrt{x}}$		\sqrt{x}		√ I ·	$\sqrt{}$	$\sqrt{}$	X	$\sqrt{1}$	/ 1 √	X	X	X	X	X	X	X	X	X
12 √ x x	$\sqrt{\mathbf{x}}$ x	$\sqrt{\mathbf{x}}$ x		√ x		√ ·	$\overline{\vee}$	$\sqrt{}$	X	√ \	/ \ \		X	X	X	X	X	X	X	X
$13 \times \sqrt{x}$	$x \sqrt{x}$	$X \sqrt{X}$		$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	<u> </u>	$\sqrt{1}$	/			X	X	X	X	X	X	X
14 x x √	$X X \sqrt{}$	X X X	X	X			X	Χ -	<u> </u>	X X		X	$\sqrt{}$	$\sqrt{}$	X	X	X	X	X	X
15 X X X	X X X	$X X \sqrt{}$	X	X √	X	X	X	X 1	<u> </u>	X >	X	X	V	<u> </u>	<u> </u>	X	X	X	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$	$\sqrt{}$ $\sqrt{}$ $\sqrt{}$ $\sqrt{}$ $\sqrt{}$		√ X	<u> </u>	√ '	<u>V</u>	<u>v</u> ,	X	$\frac{\sqrt{1}}{\sqrt{1}}$	/ 1	$\sqrt{}$	X	$\sqrt{}$	$\sqrt{}$	1	X /	X	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	V V V	$\begin{array}{c cccc} \sqrt{} & X & X \\ \hline X & \sqrt{} & X \end{array}$	1/	\sqrt{X}	1/	\\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\	<u>v_</u>	1/	X ./	$\frac{\sqrt{1}}{\sqrt{1}}$	/ V / . /	<u>\</u>	X /	<u>\</u>	<u>\</u>	1/	1/	X /	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	X X X	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	X	$\frac{V}{X} \frac{V}{}$	X	X	V X	V 1	<u>v</u>	V V X X	/ V X X	X	V 1/	X	X	X	X	V √	1/	X
20 x x x	X X X	X X X	X	x √			X	X 1	<u>*</u> /	ХУ	-#	X		X	X	X	X	V	$\sqrt{}$	
21 \ \sqrt{1} \ \sqrt{1} \ \ \sqrt{1}	x x x	х х х	X	X X	X	X	X	X	X	ХУ	ХХ	X	X	X	X	X	X	X	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	X X X	x x x	X	X X	X	x	X	X	X	х	X	X	X	X	X	X	X	X	X	X
$23 \times \sqrt{}$	X X X	X X X	X	X X	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
$24 \sqrt{x} \sqrt{x}$	X X X	X X X	X	X X	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	X X X	X X X	X	X X			X			X >		X	X	X	X	X	X	X	X	X
$26 \sqrt{x} \sqrt{x}$	X X X	XXX	X	X X		-	X			X 3	-	X	X	X	X	X	X	X	X	X
27 x x x x 28 x x x x	$\begin{array}{c cccc} V & V & V \\ X & X & V \end{array}$	X X X X	X	X X X	X	-#	X X		X X	X X X	-	X	X	X	X	X	X	X	X	X
29 x x x	$\begin{array}{c c} X & X & V \\ \hline X & \sqrt{} & \sqrt{} \end{array}$	X X X	X	X X			X			X 3	-	X	X	X	X	X	X	X	X	X
30 x x x	$\sqrt{\mathbf{x}}$	X X X	X	X X			X		X	ХУ		X	X	X	X	X	X	X	X	X
31 x x x	$X \sqrt{}$	X X X	X	X X	X	х	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
32 x x x	\sqrt{x}	X X X	X	X X	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
33 x x x	X X X	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$	X	X X	X	X	X	X	X	х х	X	X	X	X	X	X	X	X	X	X
34 x x x	X X X	X X \(\)	X	X X			X			X 3		X	X	X	X	X	X	X	X	X
35 x x x x 36 x x x x	X X X	$X \bigvee V$	X	X X	X		X		X	X X	-	X	X	X	X	X	X	X	X	X
36 x x x x 37 x x x x	X X X X X	$\begin{array}{c cccc} \sqrt{X} & \sqrt{X} & \sqrt{X} \\ \hline X & \sqrt{X} & \sqrt{X} \end{array}$	X	x x x x			X X			X X X		X	X	X	X	X	X	X	X	X
38 x x x	X X X	$\sqrt{\mathbf{x}}$	X	X X			X			ХУ		X	X	X	X	X	X	X	X	X
39 x x x	X X X	X X X	X	X X	1/	1/	X	X	X	ΧУ	X	X	X	X	X	X	X	X	X	X
40 x x x	x x x	x x x	X	\sqrt{x}	X	*/				х	-#	X	X	X	X	X	X	X	X	X
41 x x x	X X X	X X X	X	X X	X	X ·	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	$\sqrt{1}$	/ $$									
42 x x x	X X X	X X X	X	X X	X	Χ .	<u> </u>	$\sqrt{}$	<u> </u>	<u>√ </u>	/ 1			$\sqrt{}$	$\sqrt{}$		$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$
43 x x x	X X X	X X X	X	X X	X	Χ .	<u> </u>	$\sqrt{}$	$\sqrt{}$	$\frac{\sqrt{1}}{\sqrt{1}}$	<u>/ √</u>								$\sqrt{}$	$\sqrt{}$
44 x x x	X X X	X X X	X	X X				-	X	<u>√ </u>	/ X	X	X	X	X	X	X	X	X	X
45 x x x x 46 x x x	X X X X X	X X X X X	X	X X X			X X			$\frac{X}{X}$ $\frac{\chi}{X}$		X	X	X	X	X	X	X . /	X	X
47 x x x	X X X	X X X	X	X X			X			X 2	* /	$\frac{}{}$	$\frac{V}{}$	$\frac{}{}$	$\frac{}{}$	$\sqrt{}$	$\frac{V}{}$	1/	$\sqrt{}$	$\frac{V}{}$
48 x x x	X X X	X X X	X	X X						X 3	'	$\frac{v}{}$	$\frac{\mathbf{v}}{}$	$\frac{\mathbf{v}}{}$	$\frac{\mathbf{v}}{}$	v √	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	V
49 x x x	X X X	X X X	X	X X		- #				ХУ		X	X		√ √	X	X	X	X	X
50 x x x	X X X	X X X	X	X X						х			X	X	$\sqrt{}$	X	X	X	X	X
51 x x x	X X X	X X X	X	X X	X	X	X	X	X	X Y	X	X	X	X	X					
52 x x x	X X X	X X X	X	X X			X			Х У		X	X	X	X		$\sqrt{}$	$\sqrt{}$	$\sqrt{}$	$\sqrt{}$
53 x x x	X X X	X X X	X	X X		-				х х	-	X	X	X	X			$\sqrt{}$	$\sqrt{}$	$\sqrt{}$
54 x x x	X X X	XXX	X	X X						X X		X	X	X	X	X	X./	X	V	<u>V</u>
55 x x x	X X X	X X X	X	X X	X	X	X	X	X	х х	X	X	X	X	X	X	V	X	X	V

Table 8: $P(\phi)$ for $\phi = (x_1 \lor x_2 \lor x_3) \land (\overline{x}_1 \lor \overline{x}_2 \lor x_3) \land (\overline{x}_2 \lor \overline{x}_3 \lor x_4)$ and the coding example in Section 4.