

Accuracy Limitations of the Locally One-Dimensional FDTD Technique

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Abstract—While the alternating-direction implicit finite-difference time-domain (ADI-FDTD) method preserves the second-order temporal accuracy of the conventional FDTD technique, the locally one-dimensional (LOD)-FDTD method exhibits a first-order in time splitting error. Despite this difference, the numerical dispersion analyses of these methods reveal that both present similar accuracy properties. For this reason, the characteristic non-commutativity error of the LOD-FDTD scheme has not received much attention. In this work, we determine the closed form of the local truncation error for the 3D-LOD-FDTD scheme. We find that it presents error terms which depend on the time-step size multiplied by the spatial derivatives of the fields. Numerical results confirm that these terms become a significant source of error which is not revealed in the dispersion analyses.

Index Terms—Alternating direction implicit split-step FDTD methods, local truncation error, locally one-dimensional FDTD method, numerical dispersion.

I. INTRODUCTION

The finite-difference time-domain (FDTD) method is a widely used tool in computational electrodynamics. However, it is a conditionally stable technique since the maximum time-step size is limited by the Courant condition. For this reason, in problems where the size of the geometrical features is much smaller than the wavelength, the Courant condition enforces small time-step sizes that over-sample the signal, making the FDTD method computationally inefficient.

Implicit techniques, such as the Crank Nicolson (CN)-FDTD scheme, remove the Courant constraint [1]. Nevertheless, although the CN scheme presents a high degree of accuracy, it requires an extremely large sparse matrix to be solved, making the CN approach barely usable for practical problems. By contrast, Split-Step (SS)-FDTD techniques, such as the alternating-direction implicit (ADI)- and the locally one-dimensional (LOD)-FDTD methods [2]–[6], exhibit tridiagonal linear systems meeting the requirements of unconditional stability at an acceptable computational cost.

The ADI-FDTD scheme preserves the second-order temporal accuracy. Indeed, it can be viewed as an $\mathcal{O}(\Delta_t^2)$ perturbation of the CN-FDTD method [7]. The LOD-FDTD scheme is more efficient than the ADI-FDTD due to its simpler updating procedures, although it presents an $\mathcal{O}(\Delta_t)$ splitting error [1].

Although the two approaches present different accuracies in time, both lead to the same numerical dispersion relation

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[6], [9], [10]. As a result, in the framework of plane waves, both schemes yield similar numerical results. For this reason, the LOD-FDTD scheme is frequently claimed to provide the same accuracy with less computational expenditure than the ADI-FDTD technique [5], [11].

In this work, we examine in detail in what scenario the first-order splitting error of the LOD-FDTD scheme plays a critical role and leads to a significant loss in accuracy. To this end, in section II, we review the formulation and the accuracy of the CN- and the ADI-FDTD methods [7]. We then calculate the local truncation error of the 3D-LOD-FDTD scheme and compare it to the corresponding errors of the CN- and the ADI-FDTD approaches. We find that the LOD-FDTD method exhibits extra error terms that are not present in ADI- or CN-FDTD. These problematic terms are proportional to the time-step size and to the magnitude of spatial derivatives. Moreover, we show how for the two dimensional (2D) case the LOD-FDTD scheme can be expressed as an $\mathcal{O}(\Delta_t)$ perturbation of the CN-FDTD method. Finally, in section III, we illustrate the influence of the aforementioned errors, simulating a WR75 waveguide cavity.

II. FORMULATION AND ACCURACY ANALYSIS

Three dimensional (3D) Maxwell curl equations in lossless isotropic media can be expressed as

$$\frac{\partial \vec{\mathcal{U}}}{\partial t} - [\mathcal{R}] \vec{\mathcal{U}} = 0 \quad (1)$$

where $\vec{\mathcal{U}}$ is the composite field vector given by

$$\vec{\mathcal{U}} = (\mathcal{E}_x, \mathcal{E}_y, \mathcal{E}_z, \mathcal{H}_x, \mathcal{H}_y, \mathcal{H}_z)^T \quad (2)$$

and the superscript T denotes matrix transpose. The matrix $[\mathcal{R}]$ can be written as the sum of two operators

$$[\mathcal{R}] = [\mathcal{A}] + [\mathcal{B}]$$

where $[\mathcal{A}]$ and $[\mathcal{B}]$ are 6×6 block matrices defined as

$$[\mathcal{A}] = \begin{pmatrix} 0 & \frac{1}{\epsilon}[\mathcal{R}_e] \\ -\frac{1}{\mu}[\mathcal{R}_o] & 0 \end{pmatrix}, \quad [\mathcal{B}] = \begin{pmatrix} 0 & \frac{1}{\epsilon}[\mathcal{R}_o] \\ -\frac{1}{\mu}[\mathcal{R}_e] & 0 \end{pmatrix} \quad (3)$$

with $[\mathcal{R}_e]$ and $[\mathcal{R}_o]$ given by

$$[\mathcal{R}_e] = \begin{pmatrix} 0 & 0 & \frac{\partial}{\partial y} \\ \frac{\partial}{\partial z} & 0 & 0 \\ 0 & \frac{\partial}{\partial x} & 0 \end{pmatrix}, \quad [\mathcal{R}_o] = \begin{pmatrix} 0 & -\frac{\partial}{\partial z} & 0 \\ 0 & 0 & -\frac{\partial}{\partial x} \\ -\frac{\partial}{\partial y} & 0 & 0 \end{pmatrix}.$$

Thus, the LOD-FDTD scheme introduces additional first-order error terms that depend on the time-step size and the spatial derivatives of the fields. It will be shown in the results section that these error terms will be dominant in those regions with strong spatial variation of the fields, e. g. around singularities associated with corners or near-field sources.

2) *Two dimensional case:* For the TE_z polarization $\vec{U} = (\mathcal{E}_x, \mathcal{E}_y, \mathcal{H}_z)^T$, and for the TM_z case $\vec{U} = (\mathcal{E}_z, \mathcal{H}_x, \mathcal{H}_y)^T$. In both cases, the matrix $[\mathcal{R}]$ is reduced to a 3×3 matrix that can be decomposed as the sum of two operators, $[\mathcal{A}]$ and $[\mathcal{B}]$. These operators can be obtained particularizing (3) for the TE_z and the TM_z polarizations.

It can be seen that, for the TE_z and the TM_z cases, the following equality holds

$$(\mathbf{AB} - \mathbf{BA}) \vec{U}^{n+\frac{1}{2}} = \mathbf{AB} \vec{U}^n - \mathbf{BA} \vec{U}^{n+1} \quad (17)$$

where \mathbf{A} , \mathbf{B} and \vec{U} are the numerical counterpart of $[\mathcal{A}]$, $[\mathcal{B}]$ and \vec{U} for each polarization. Substituting (17) into (14) we thus obtain

$$\frac{\vec{U}^{n+1} - \vec{U}^n}{\Delta_t} - \left[\mathbf{R} - \frac{\Delta_t}{2} (\mathbf{AB} - \mathbf{BA}) \right] \frac{\vec{U}^{n+1} + \vec{U}^n}{2} = 0. \quad (18)$$

Comparing this expression to (4), it is straightforward to see that the 2D-LOD-FDTD scheme can be recognized as an $\mathcal{O}(\Delta_t)$ perturbation of the CN-FDTD approach.

a) *TE_z-LOD-FDTD:* Substituting the exact field solution \vec{U} of (1) into (18) and using the Taylor series we obtain the truncation error of the TE_z -LOD-FDTD scheme

$$\vec{T}_{\text{LOD-TE}_z} = (\mathbf{E}_{\text{CN-TE}_z} + \mathbf{E}_{\text{LOD-TE}_z}) \vec{U}(\vec{r}_m, t_m) \quad (19)$$

where $\mathbf{E}_{\text{CN-TE}_z}$ can be obtained particularizing (5) for the TE_z case and $\mathbf{E}_{\text{LOD-TE}_z}$ is the matrix that contains the error terms unique to the TE_z -LOD-FDTD scheme. Thus, the local truncation error for the TE_z case can be written as

$$\vec{T}_{\text{LOD-TE}_z} = \begin{pmatrix} 2h_t & \boxed{f_{xy}} & \frac{h_y}{\epsilon} \\ \boxed{-f_{xy}} & 2h_t & -\frac{h_x}{\epsilon} \\ \frac{h_y}{\mu} & -\frac{h_x}{\mu} & 2h_t \end{pmatrix} \vec{U}(\vec{r}_m, t_m) \quad (20)$$

where h_u was given in equation (6) and $\vec{U}(\vec{r}_m, t_m) = (\mathcal{E}_x(\vec{r}_1, t_1), \mathcal{E}_y(\vec{r}_2, t_2), \mathcal{H}_z(\vec{r}_3, t_3))^T$ with $t_m \in (n\Delta_t, (n+1)\Delta_t)$ and \vec{r}_m is inside an ellipse of center $(i\Delta_x, j\Delta_y)$ and semi-axes $(\Delta_x/2, \Delta_y/2)$.

In (20), the boxed terms correspond to the $\mathbf{E}_{\text{LOD-TE}_z}$ matrix. When those terms are set to zero, the matrix in (20) is reduced to $\mathbf{E}_{\text{CN-TE}_z}$. It should be noted that, unlike \mathbf{E}_{LOD} , the matrix $\mathbf{E}_{\text{LOD-TE}_z}$ only contains first-order error terms in time. This result is in agreement with (18).

b) *TM_z-LOD-FDTD:* For the TM_z polarization the local truncation error can be expressed as

$$\vec{T}_{\text{LOD-TM}_z} = (\mathbf{E}_{\text{CN-TM}_z} + \mathbf{E}_{\text{LOD-TM}_z}) \vec{U}(\vec{r}_m, t_m) \quad (21)$$

where $\mathbf{E}_{\text{CN-TM}_z}$ can be obtained by particularizing (5) for the TM_z case and $\mathbf{E}_{\text{LOD-TM}_z}$ contains the error terms unique to the TM_z -LOD-FDTD scheme. The error for the TM_z case

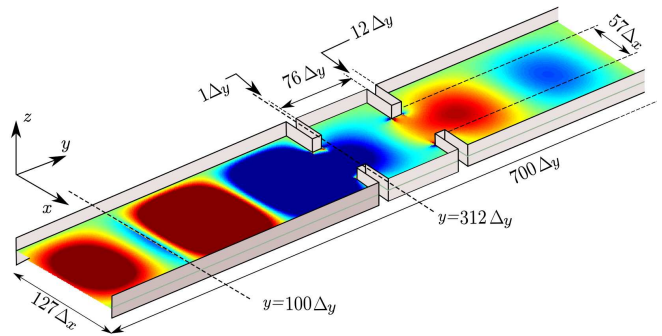


Fig. 1. Geometry of the problem simulated. Snapshot of the electric field computed by means of the LOD-FDTD technique in part of the simulation domain ($N_\lambda = 205.77$, $N_t = 45.2$).

can thus be written as [13]

$$\vec{T}_{\text{LOD-TM}_z} = \begin{pmatrix} 2h_t & -\frac{h_y}{\epsilon} & \frac{h_x}{\epsilon} \\ -\frac{h_y}{\mu} & 2h_t & \boxed{-f_{xy}} \\ \frac{h_x}{\mu} & \boxed{f_{xy}} & 2h_t \end{pmatrix} \vec{U}(\vec{r}_m, t_m) \quad (22)$$

where h_u was given in equation (6) and $\vec{U}(\vec{r}_m, t_m) = (\mathcal{E}_z(\vec{r}_1, t_1), \mathcal{H}_x(\vec{r}_2, t_2), \mathcal{H}_y(\vec{r}_3, t_3))^T$ with $t_m \in (n\Delta_t, (n+1)\Delta_t)$ and \vec{r}_m is inside an ellipse of center $(i\Delta_x, j\Delta_y)$ and semi-axes $(\Delta_x/2, \Delta_y/2)$.

The boxed terms correspond to the $\mathbf{E}_{\text{LOD-TM}_z}$ matrix. Again, unlike \mathbf{E}_{LOD} , the matrix $\mathbf{E}_{\text{LOD-TM}_z}$ only contains first-order error terms in time, since the 2D-LOD-FDTD algorithm can be seen as a first-order perturbation of the CN-FDTD.

III. NUMERICAL RESULTS

To illustrate the influence of the error terms calculated in the previous section we have considered a WR75 waveguide cavity. The geometry of the problem is shown in Fig. 1. To simulate the propagation of the TE_{10} mode we have introduced a modal source located at $y = 100\Delta_y$. The source was a modulated Gaussian pulse with central frequency $f_0 = 12.5$ GHz and effective lateral bandwidth $\text{BW} = 2.5$ GHz. The cell size was $\Delta_x = \Delta_y = \Delta = 0.15$ mm and the spatial resolution was $N_\lambda = \lambda_g/\Delta = 205.77$, with λ_g being the wavelength in the waveguide at the central frequency f_0 . The ports of the waveguide were terminated by second-order one-way wave equation ABCs based on the Higdon operator [12].

We have simulated the propagation of the fundamental mode and computed the resonant frequency. To estimate the error, we have considered the solution obtained with the conventional FDTD method as a benchmark. In this case, the cell size was $\Delta/4 = 0.0375$ mm and the time step was the maximum allowed by the stability condition $\Delta_{t_0} = 0.08845$ ps. The resonant frequency of the dominant mode computed by means of the FDTD simulation was $f = 12.789$ GHz. Fig. 2 shows the absolute error of the results obtained with ADI and LOD as a function of the temporal resolution $N_t = T_0/\Delta_t$, being T_0 the wave period at the frequency f_0 . We can see how both approaches provide practically the same solution. The slope of the curves confirms second-order convergence in time of the solution. The overall accuracy reached is limited by the value of the spatial resolution N_λ that controls the spatial error.

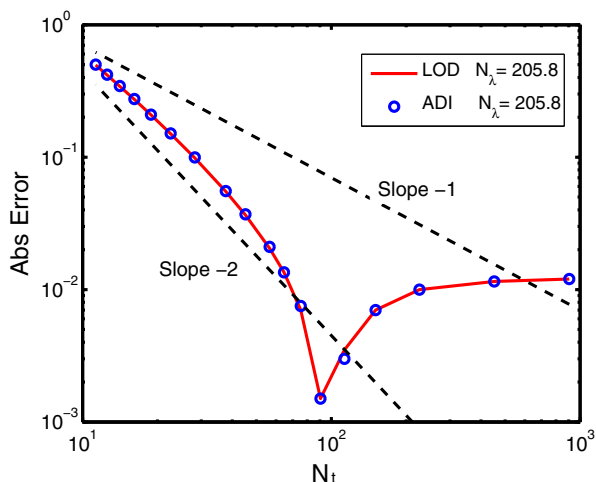


Fig. 2. Absolute error of the resonant frequency as a function of N_t . ADI and LOD provide the same solution. Convergence follows a second-order rate in time. The spatial resolution $N_\lambda = 205.77$ limits the accuracy achieved.

Although not shown explicitly here, the ADI and the LOD techniques yield the same scattering parameters in the whole frequency band, from 10 GHz to 15 GHz. This result is linked to the fact that both schemes present the same numerical dispersion relation [6], [9]. However, as indicated by the error analysis presented in section II, the ADI-FDTD scheme is second-order accurate in time while the LOD-FDTD method is only first-order accurate. Thus, in order to show the influence of the first-order error terms obtained in (22), we have computed the field pattern of the E_z component at the plane $y = 312 \Delta_y$ (see Fig. 1). The field pattern calculated by using the LOD-FDTD, the ADI-FDTD and the conventional FDTD schemes are shown in Fig. 3. It can be seen that, near the iris edges, where the fields present a strong spatial variation, the LOD-FDTD scheme leads to major errors. However, as we move away from the irises, and for a given value of N_λ and N_t , the field values given by the ADI- and the LOD-FDTD coincide. The degradation of the LOD-FDTD solution near the iris corners can also be seen in the snapshot shown in Fig. 1.

IV. CONCLUSION

The LOD-FDTD method achieves better CPU-time efficiency than the ADI-FDTD. In addition, numerical dispersion analyses reveal that it provides similar results to the ADI-FDTD. Nevertheless, the LOD-FDTD scheme presents an $\mathcal{O}(\Delta_t)$ splitting error. In this work, the closed form of the local truncation error of the 3D-LOD-FDTD scheme is given. We find that it exhibits first-order in time error terms that depend on the time-step size multiplied by the spatial derivatives of the fields. These terms, which are not present in the ADI- or the CN-FDTD schemes, become a significant error source that does not show up in dispersion analysis. Moreover, for the two dimensional case, we find that the LOD-FDTD method can be expressed as an $\mathcal{O}(\Delta_t)$ perturbation of the CN-FDTD scheme.

Finally, we have simulated the propagation of the fundamental mode in a WR75 waveguide cavity. The results confirm that these error terms are critical when the fields

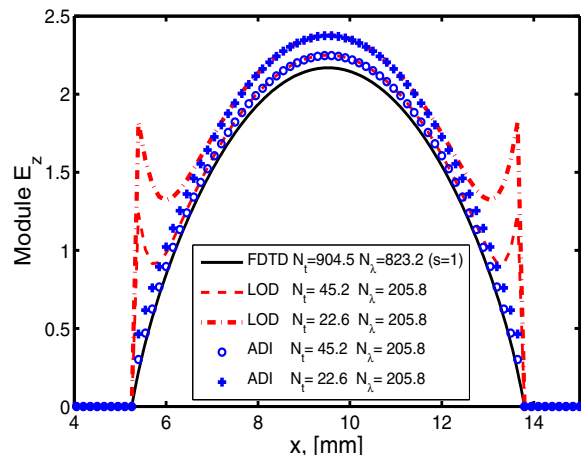


Fig. 3. Field pattern of the E_z component at the plane $y = 312\Delta_y$ calculated by the LOD-FDTD, the ADI-FDTD and the conventional FDTD schemes.

present strong spatial variations. Hence, statements such as “The computational time is reduced with the accuracy being maintained, when compared with the ADI-FDTD...” made in [5] when talking about the LOD-FDTD method should be framed in an appropriate context.

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