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Frequency response of $LaAlO_3/SrTiO_3$ all-oxide field-effect transistors

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ABSTRACT

The frequency response of all oxide field-effect transistors with amorphous $LaAlO_3$ on a crystalline $SrTiO_3$ substrate is reported. The intrinsic cut-off frequencies of 4 μm gate-length devices are found to be approximately 17 MHz indicating that with gate length scaling gigahertz cut-off frequency is possible. The low cut-off frequency is primarily limited by the low effective mobility. The estimated effective mobility value determined from the S-parameter measurement is 3.8 cm^2/Vs, which is consistent with previous reports. Small-signal equivalent circuit model parameters are extracted by fitting to on-wafer measured S-parameters. Good agreement is obtained between measured and simulated S-parameters based on the equivalent circuit model.

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1. Introduction

Since Ohtomo and Hwang reported a high-mobility two-dimensional electron gas (2DEG) can be formed at the crystalline $LaAlO_3/SrTiO_3$ heterointerface in 2004 [1], this 2DEG has received increasing attention [2–7]. Dong et al. demonstrated the first all-oxide FETs based on modulation of the 2DEG at the interface of amorphous $LaAlO_3$ and crystalline $SrTiO_3$ [8,9]. These reported transistors showed an on-current of 10 μA/μm with an on–off ratio around 1000, and exhibited gate leakage below 10 fA/μm^2. The room temperature mobility was estimated to be approximately 4 cm^2/Vs [9], which can be expected to limit the speed of these transistors. However, until now no direct measurements of the frequency response of all-oxide FETs have been reported. The application space for all-oxide transistors is still to be determined.

2. Experiment

The measured $LaAlO_3/SrTiO_3$ FETs in this paper incorporated a 3 nm Al_2O_3 capping layer on top of 8 nm $LaAlO_3$, both of which were deposited by atomic layer deposition (ALD) on a crystalline $SrTiO_3$ substrate. The schematic cross section and a microscope image of the transistor is shown in Fig. 1. The details of the fabrication process are reported in [9]. Three sets of devices fabricated using slightly different $LaAlO_3$ deposition conditions were characterized. Sample A was initiated with La in the first cycle of an 8 nm $LaAlO_3$ film; sample B was the same except the first cycle was Al; and in sample C the $SrTiO_3$ substrate was annealed in O_2 at 1000 °C for 10 min prior to the La-first-cycle 8 nm $LaAlO_3$ deposition.

The frequency response of the transistors was characterized by on-wafer S-parameter measurements using 150-μm pitch ground-signal-ground (GSG) coplanar probes. The S-parameters were measured from 30 kHz to 30 MHz using an Agilent 8753E vector network analyzer. An open-short-load-through calibration was performed using a CS-5 calibration substrate from GGB Industries to establish reference planes at the probe tips. The dc bias was supplied by an Agilent 4155C semiconductor parameter analyzer connected through network analyzer bias tees. Current voltage (I–V) measurements were performed through the GSG probes on the same S-parameter test structure. Measured common source characteristics are shown in Fig. 2. The dc I–V characteristics are similar to those reported in [9].

3. Results and discussion

The current gain, h_{21}, and the unilateral power gain, U, of typical 4 μm gate length transistors for each of the three $LaAlO_3/SrTiO_3$ interface conditions is shown in Fig. 3, where all transistors are biased at their peak current gain cut-off frequency, f_T. Relatively modest differences are observed for the frequency response among these three samples, with the oxygen-treated sample performing

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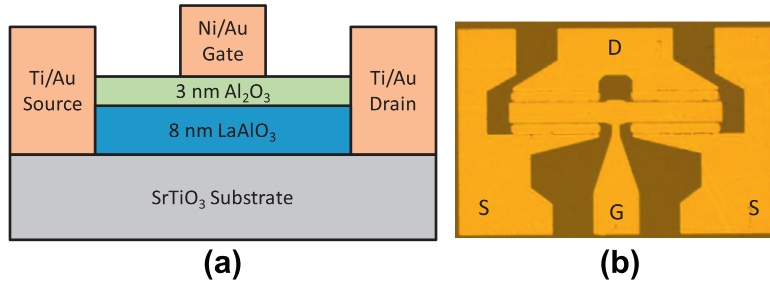

Fig. 1. (a) Schematic cross section and (b) a microscope image of the $LaAlO_3/SrTiO_3$ all-oxide FET.

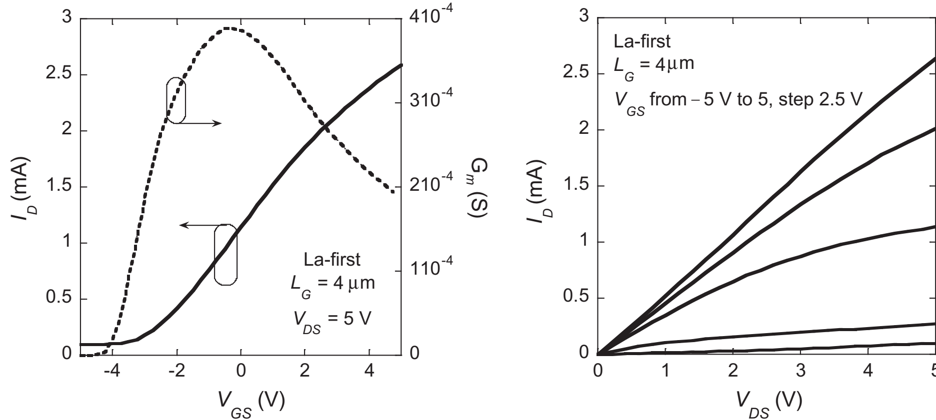

Fig. 2. Measured (a) drain current vs. gate-source voltage, I_D–V_{GS}, and (b) drain current vs. drain–source voltage, I_D–V_{DS}, for a representative $LaAlO_3/SrTiO_3$ FET.

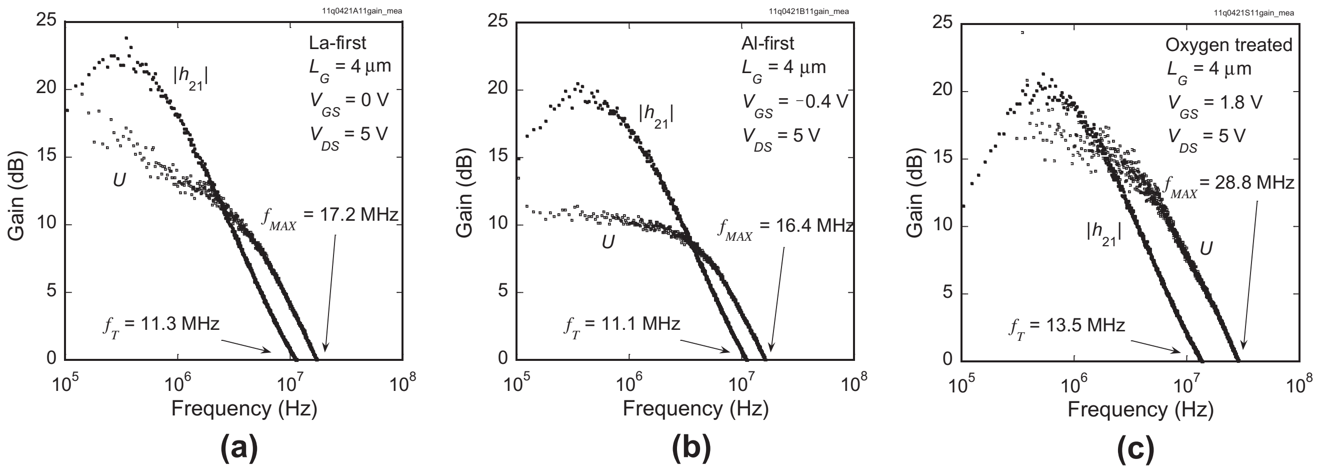

Fig. 3. Small-signal RF performance of the transistors with 4-µm gate length on three samples: (a) La-first, (b) Al-first, and (c) oxygen treated.

slightly better than either of the un-annealed samples. The low cut-off frequency and maximum oscillation frequency of the transistors is primary due to the low effective channel mobility, which leads to low transconductance, g_m, and high source and drain access resistances.

In order to investigate the transistors' intrinsic frequency response, a small-signal equivalent circuit model was constructed and the extrinsic parasitics were extracted following the procedure outlined in [10]. The small-signal equivalent circuit model is shown in Fig. 4; in this conventional transistor model, R_G, R_S, and R_D represent the gate, source, and drain resistances, and C_{GS}, C_{GD}, and C_{DS} represent the gate-source, gate-drain, and drain–source capacitances. r_O is the small-signal output resistance. The

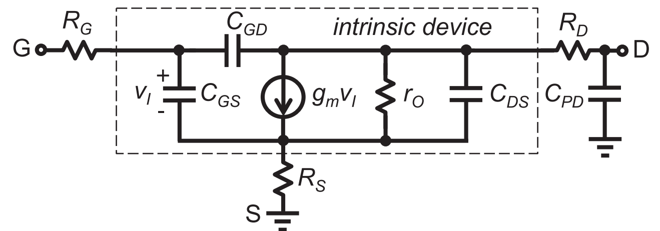

Fig. 4. Small-signal equivalent circuit of the all-oxide FET.

voltage-controlled current source is expressed in terms of the transconductance, g_m, and the voltage V_1 appearing across the

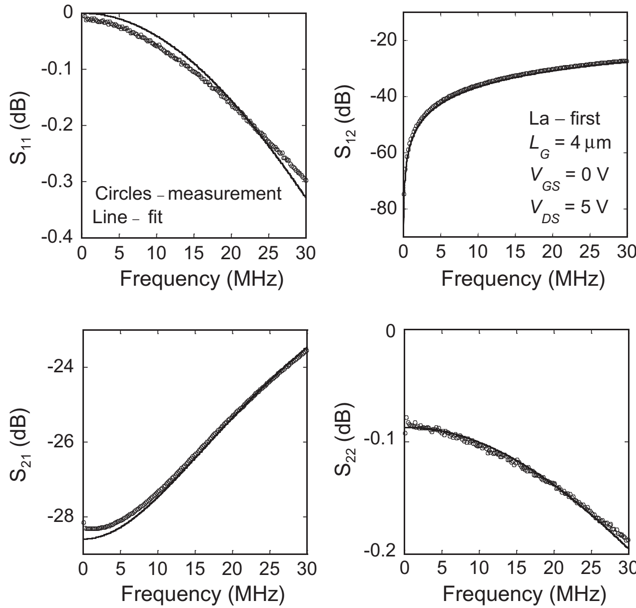

Fig. 5. Comparison of measured (line) and simulated (circles) S-parameters showing quality of fit.

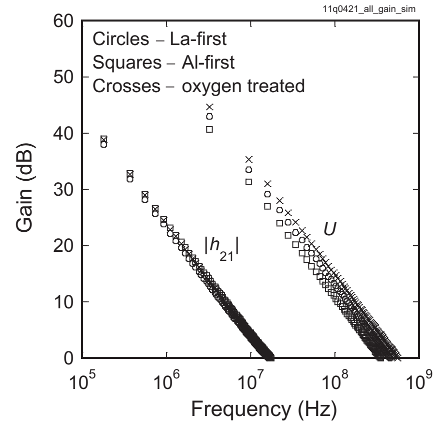

Fig. 6. Calculated the intrinsic RF performance of the transistors based on the small-signal equivalent circuit in Fig. 3 on three samples. Sample A: La-first (circles); sample B: Al-first (squares); and sample C: oxygen treated (crosses).

intrinsic gate-source capacitance. The capacitance C_{PD} represents the parasitic capacitance of the contact pads. The extracted parameter values of a typical transistor on sample A (La-first) are $R_G \approx 0\,\Omega$, $R_S = R_D = 643.5\,\Omega$, $C_{GS} = 4.34$ pF, $C_{GD} = 2.24$ pF, $g_m = 592$ µS, $r_O = 6.29$ kΩ, and $C_{PD} = 1.31$ pF; similar values were obtained for devices on the other two samples. A parasitic drain-to-source capacitance in the picofarad range is needed to account for the excess capacitance associated with the drain pad in these non-isolated devices. A similar value of parasitic capacitance was also measured in an open-circuit coplanar structure on each sample, indicating that this is a layout-induced parasitic and is not related to the intrinsic device structure.

Good agreement between measured S-parameters and the equivalent circuit model fits was obtained for all the transistors on three samples. Fig. 5 shows a typical example for a typical transistor on sample A; as can be seen, the frequency response of all of the S-parameters tracks closely the model behavior. The extracted transconductance from the small-signal equivalent circuit was compared to the derivative of the measured dc I–V curve. Good agreement between the small signal equivalent circuit model extraction and the dc measurement supported the validity of the equivalent circuit model. For example, for the three samples A, B, and C, the measured transconductance from S-parameter measurements was 0.37, 0.35, and 0.30 mS, while the transconductance from dc measurements at the same bias condition was 0.39, 0.40, and 0.35 mS.

The extracted source resistance and drain resistance are between 120 and 180 kΩ-µm at a gate width of 200 µm and the gate-source/drain spacing of 1 µm for all three samples. This extracted resistance can be accounted for by the high sheet resistance in the 2DEG, which is estimated to be in the range of 120–180 kΩ/sq, and is close to the reported values in [9]. In general, it is challenging to minimize parasitic resistance in all-oxide transistors.

By mathematically removing the effects of the extrinsic resistances, R_G, R_S, and R_D, and capacitance, C_{PD}, the intrinsic device RF performance can be projected, as shown in Fig. 6. The maximum frequency of oscillation, f_{max}, increases by a factor of ~20, due to significantly reducing the output RC time constant with estimated values of 13–15 ns for the three samples. The cut-off frequency, on the other hand, does not increase as much because it is limited primarily by the low effective mobility with estimated values of 4–6 cm^2/Vs and the long channels. With scaling the channel length below 50 nm, we may expect the cut-off frequency over 1 GHz.

To estimate the effective mobility, we employ a first-order approximation. Assuming the transistor is biased in the triode region and the charge and the electric field are uniformly distributed in the channel, the drain current can be estimated as $I_D = Qv/L = Q\mu E/L = Q\mu[V_{DS} - I_D\ (R_S + R_D)]/L^2$, where Q is the total charge in the channel, v is the velocity, L is the gate length, and μ is the effective mobility. The charge Q can be estimated by the ac measurement as $Q = C_G V_{GS}$, where C_G is the total gate capacitance. Therefore the effective mobility can be estimated as $\mu = I_D L^2/\{C_G V_{GS}[V_{DS} - I_D\ (R_S + R_D)]\}$. The effective mobility of the transistor on sample A (La-first) is estimated to be 3.8 cm^2/Vs, which is consistent to the previous reported value of 3.9 cm^2/Vs in [3]. Similar results are obtained on the other structures, with 3.8 and 5.6 cm^2/Vs for sample B and C, respectively. The effective mobility can also be extracted based on the Y-function, $Y = I_D/\sqrt{g_m}$, or peak transconductance after Ghibaudo [11]. The extracted mobilities based on these two techniques for all three wafers are in the range of 2–4 cm^2/Vs, with good agreement between extraction methods.

From the mobility and sheet resistance, the charge density in the 2DEG can be estimated as $N = 1/q\mu R_{SH}$, where R_{SH} represents the sheet resistance of the 2DEG and can be extracted from transmission line measurements. As an example, the extracted sheet resistance on sample A (La-first) is approximately 117 kΩ/sq, which is close to the reported value of 140 kΩ/sq in [9]. Therefore, the charge density is estimated to be 1.4×10^{13} cm^{-2}.

4. Conclusion

The frequency response of the $LaAlO_3/SrTiO_3$ all-oxide FETs is measured for the first time. The extracted parameters from the small-signal equivalent circuit are consistent with the dc characteristics. The effective mobility extracted from the ac measurements is also consistent with the dc measurements and confirms a value of approximately 4 cm^2/Vs. The intrinsic cut-off frequencies on 4 µm gate-length devices are found to be approximately 17 MHz indicating that with gate length scaling gigahertz cut-off frequency is possible in all-oxide transistors.

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